

Chapter 5 Circuit Theorems

Exercises

Ex 5.3-1 $R = 10 \Omega$ and $i_s = 1.2 \text{ A}$.

Ex 5.3-2 $R = 10 \Omega$ and $i_s = -1.2 \text{ A}$.

Ex 5.3-3 $R = 8 \Omega$ and $v_s = 24 \text{ V}$.

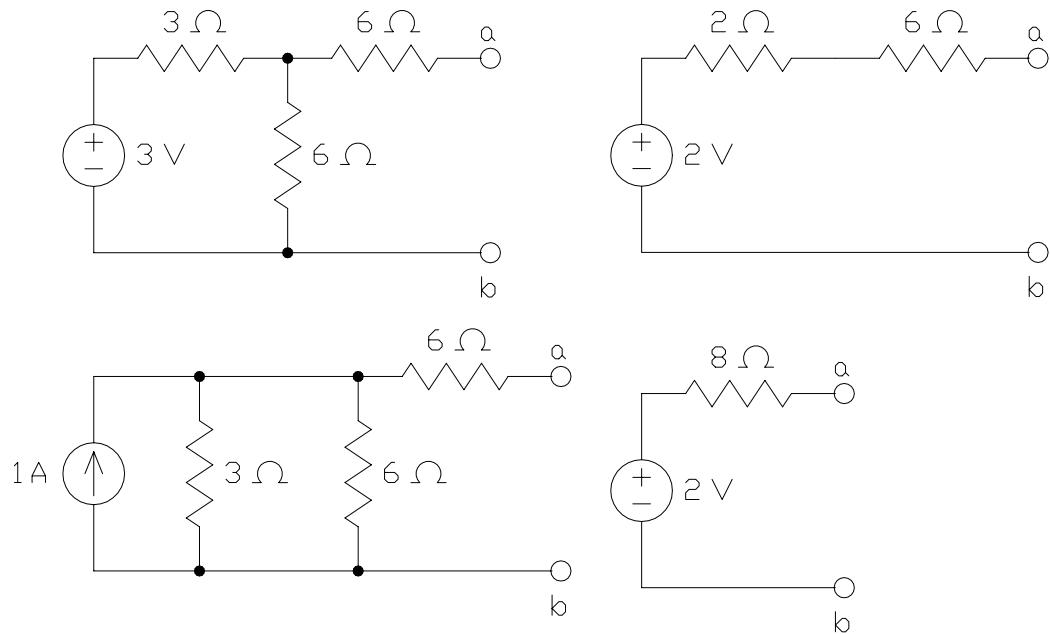
Ex 5.3-4 $R = 8 \Omega$ and $v_s = -24 \text{ V}$.

$$\text{Ex 5.4-1} \quad v_m = \frac{20}{10+20+20} 15 + 20 \left(-\frac{10}{10+(20+20)} 2 \right) = 6 + 20 \left(-\frac{2}{5} \right) = -2 \text{ V}$$

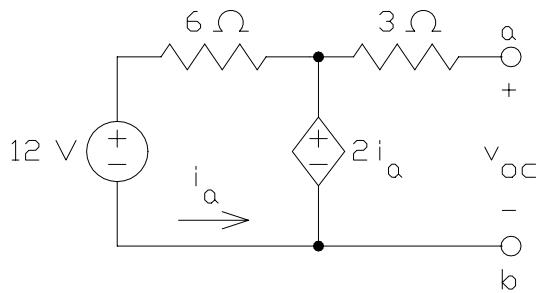
$$\text{Ex 5.4-2} \quad i_m = \frac{25}{3+2} - \frac{3}{2+3} 5 = 5 - 3 = 2 \text{ A}$$

$$\text{Ex 5.4-3} \quad v_m = 3 \left(\frac{3}{3+(3+3)} 5 \right) - \frac{3}{3+(3+3)} 18 = 5 - 6 = -1 \text{ A}$$

Ex 5.5-1

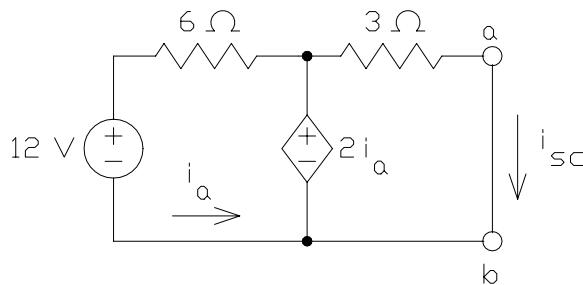


Ex 5.5-2



$$i_a = \frac{2i_a - 12}{6} \Rightarrow i_a = -3 \text{ A}$$

$$v_{oc} = 2i_a = -6 \text{ V}$$

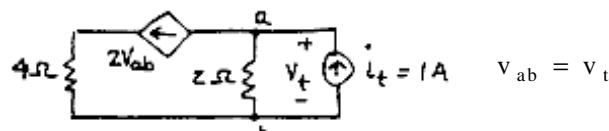


$$12 + 6i_a = 2i_a \Rightarrow i_a = -3 \text{ A}$$

$$3i_{sc} = 2i_a \Rightarrow i_{sc} = \frac{2}{3}(-3) = -2 \text{ A}$$

$$R_t = \frac{-6}{-2} = 3 \Omega$$

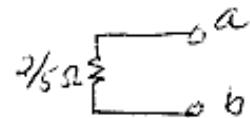
Ex. 5.5-3 No independent sources $\therefore v_{oc} = i_{sc} = 0 \Rightarrow$ apply 1A test source



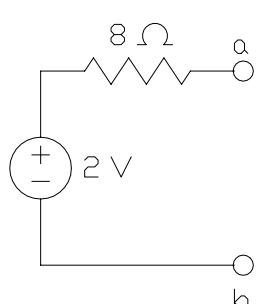
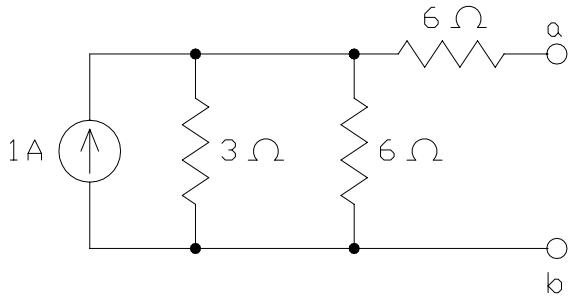
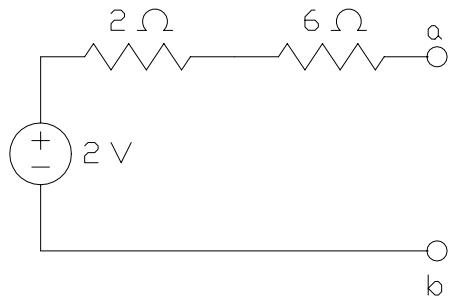
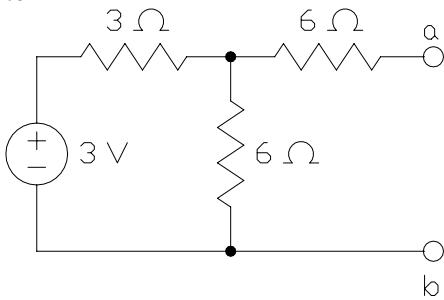
$$\text{KCL at } a: 2v_t + \frac{v_t}{2} - 1 = 0 \Rightarrow v_t = \frac{2}{5} \text{ V}$$

Thev. equiv. ckt

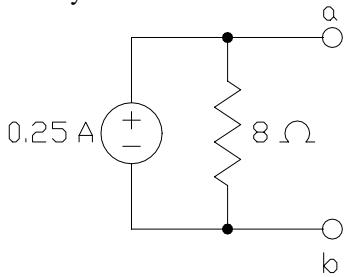
$$\therefore R_T = \frac{v_t}{i_t} = \frac{2}{5} \Omega$$



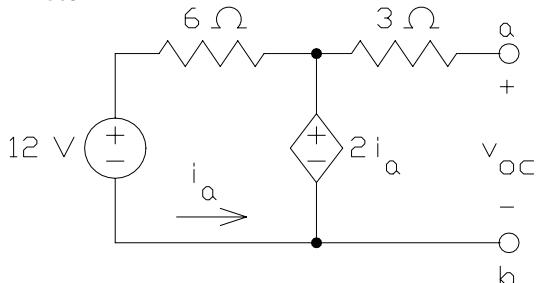
Ex 5.6-1



Finally:

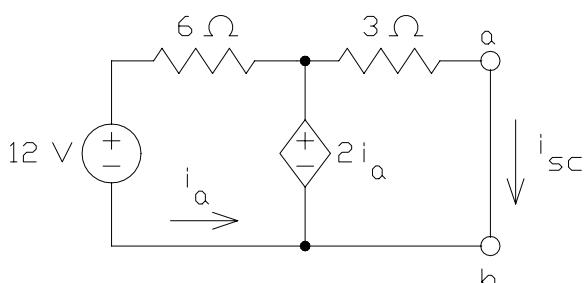


Ex 5.6-2



$$i_a = \frac{2i_a - 12}{6} \Rightarrow i_a = -3 \text{ A}$$

$$v_{oc} = 2i_a = -6 \text{ V}$$

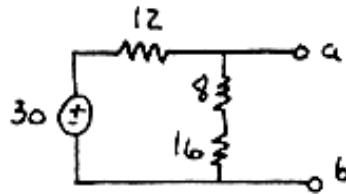


$$12 + 6i_a = 2i_a \Rightarrow i_a = -3 \text{ A}$$

$$3i_{sc} = 2i_a \Rightarrow i_{sc} = \frac{2}{3}(-3) = -2 \text{ A}$$

$$R_t = \frac{-6}{-2} = 3 \Omega$$

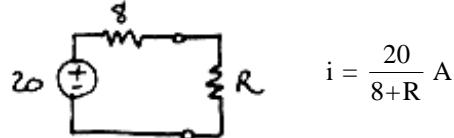
Ex. 5.6-3



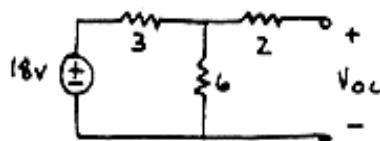
$$R_T = \frac{12 \times 24}{12+24} = \frac{12 \times 24}{36} = 8\Omega$$

$$v_{oc} = \frac{24}{12+24} 30 = 20V$$

So we have



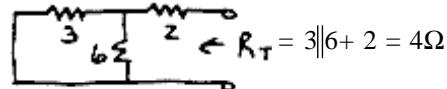
Ex. 5.7-1 Find v_{oc}



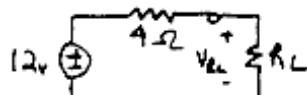
From voltage divider

$$v_{oc} = 18V \left(\frac{6}{6+3} \right) = 12V$$

Find R_T (short 18V source)

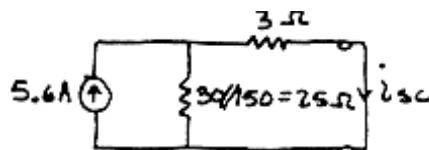


\therefore Thev. equiv ckt \Rightarrow



$$\text{For max power to } R_L \Rightarrow R_L = R_T = 4\Omega \quad \therefore P_{max} = \frac{(v_{RL})^2}{R_L} = \frac{(6)^2}{4} = 9W$$

Ex. 5.7-2 Find i_{sc}



From current divider

$$i_{sc} = 5.6A \left(\frac{25}{25+3} \right)$$

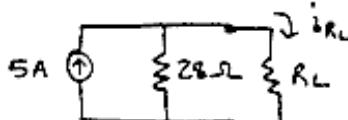
$$\underline{i_{sc} = 5A}$$

Find R_T (open 5.6A source)



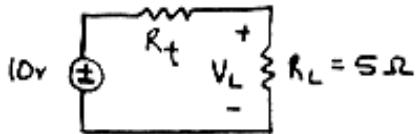
$$R_T = 25+3 = 28\Omega$$

\therefore Norton equiv. ckt \Rightarrow



$$\text{For max power } R_L = R_T = 28\Omega \quad \therefore P_{L_{max}} = (i_{R_L})^2 R_L = (5/2)^2 (28) = 175W$$

Ex. 5.7-3



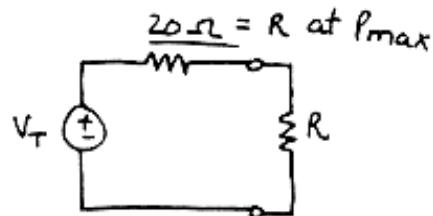
$$P_{L_{max}} = \frac{(V_{L_{max}})^2}{R_L} = \frac{\left[10V \left(\frac{5}{5+R_t}\right)\right]^2}{R_L}$$

Now for V_L to be maximized, R_t must be minimized

\therefore choose $R_t = 1\Omega$

$$\therefore P_{L_{max}} = \frac{\left[10\left(\frac{5}{6}\right)\right]^2}{5} = 13.9W$$

Ex. 5.7-4



$$P_{max} = 5 = \left(\frac{v_T}{40}\right)^2 20 = \frac{v_T^2}{80}$$

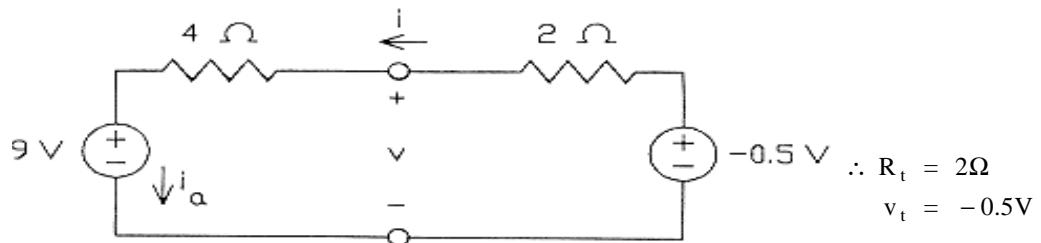
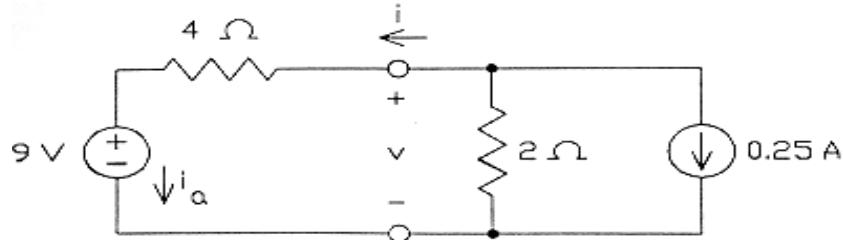
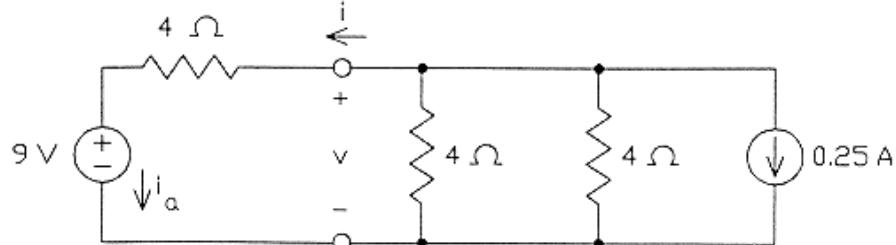
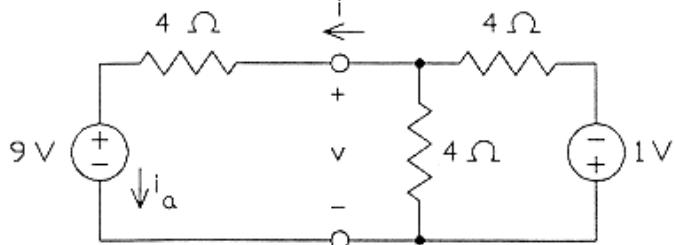
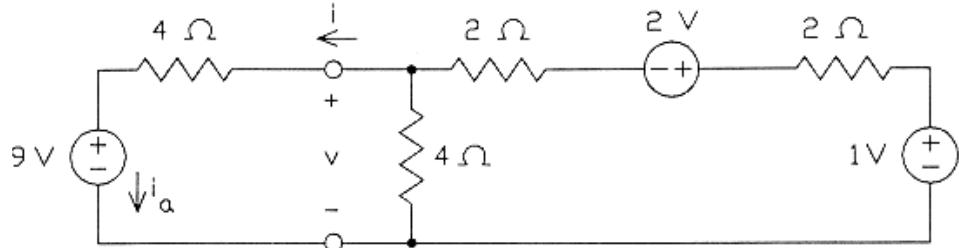
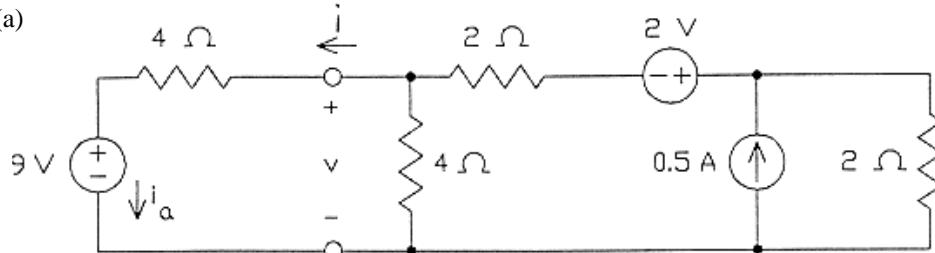
$$v_T = \sqrt{400} = 20V$$

PROBLEMS

Section 5-3: Source Transformations

P5.3-1

(a)



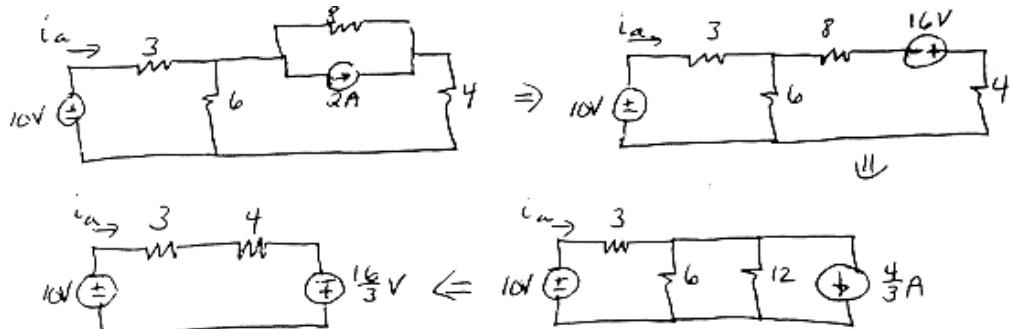
$$(b) \quad -9 - 4i - 2i + (-0.5) = 0$$

$$i = \frac{-9 + (-0.5)}{4+2} = -1.58A$$

$$v = 9 + 4i = 9 + 4(-1.58) = 2.67V$$

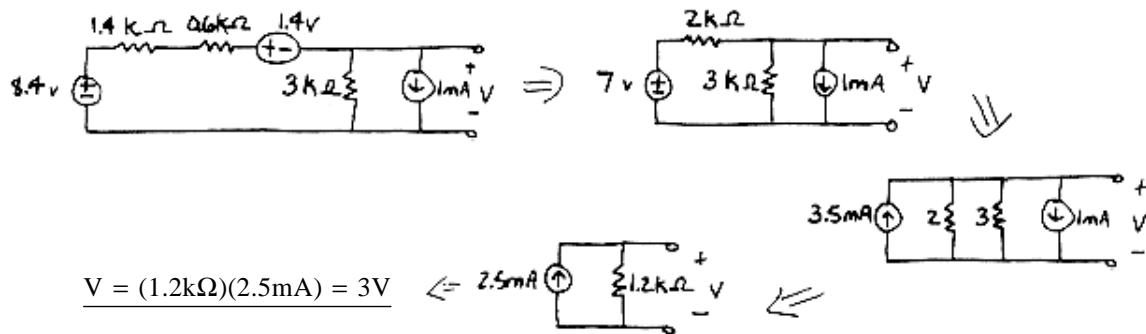
$$(c) \quad i_a = i = -1.58A$$

P5.3-2

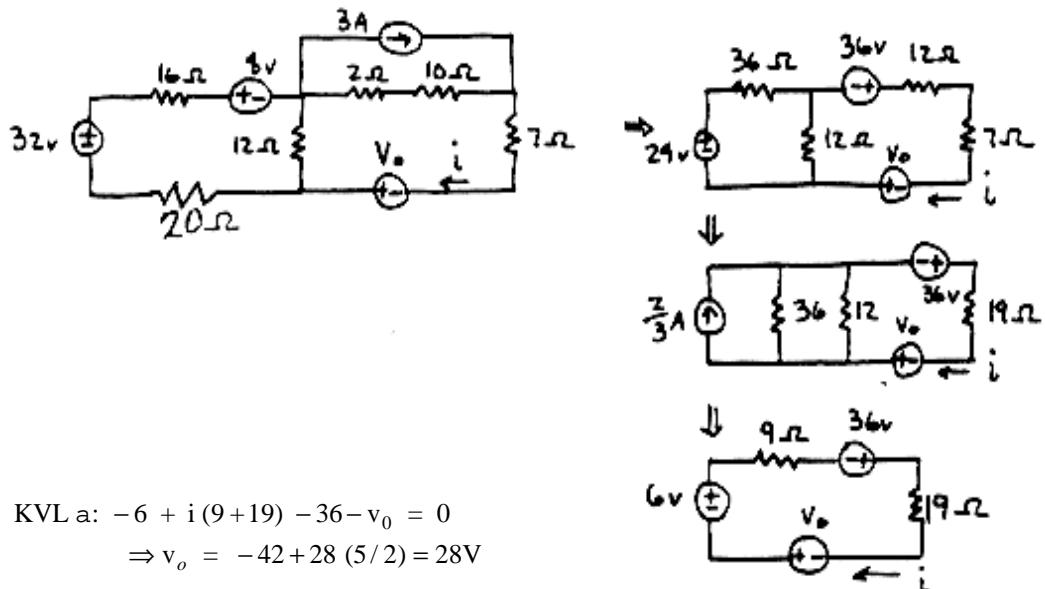


$$\text{KVL: } -10 + 3i_a + 4i_a - \frac{16}{3} = 0 \quad \therefore i_a = 2.19\text{A}$$

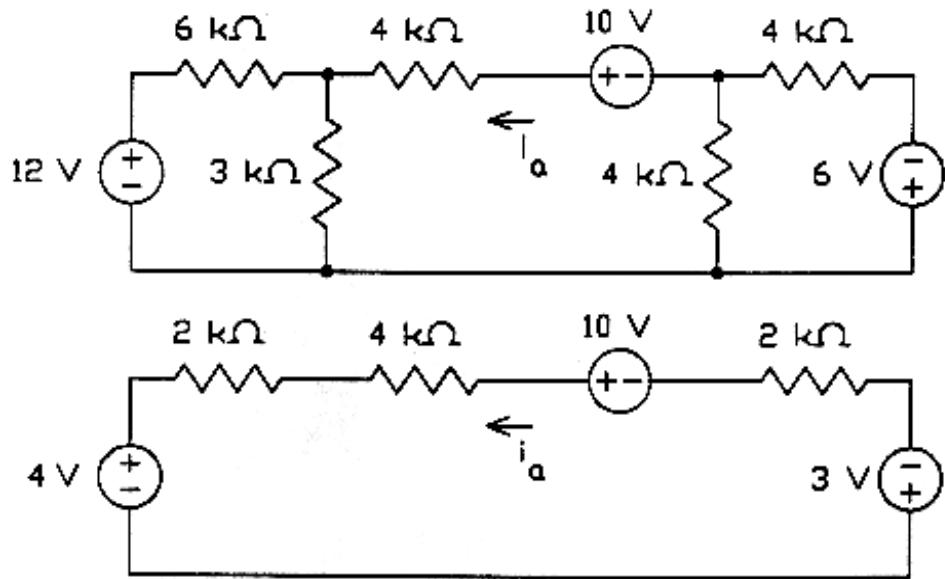
P5.3-3



P5.3-4



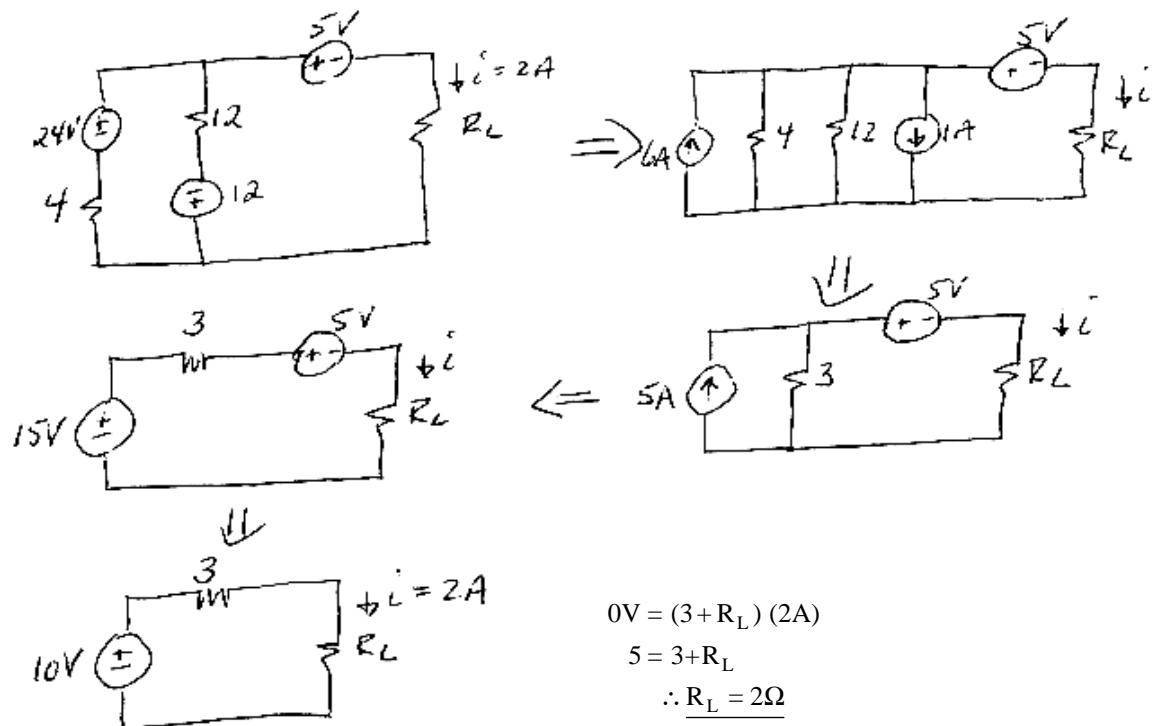
P5.3-5



$$-4 - 2000i_a - 4000i_a + 10 - 2000i_a - 3 = 0$$

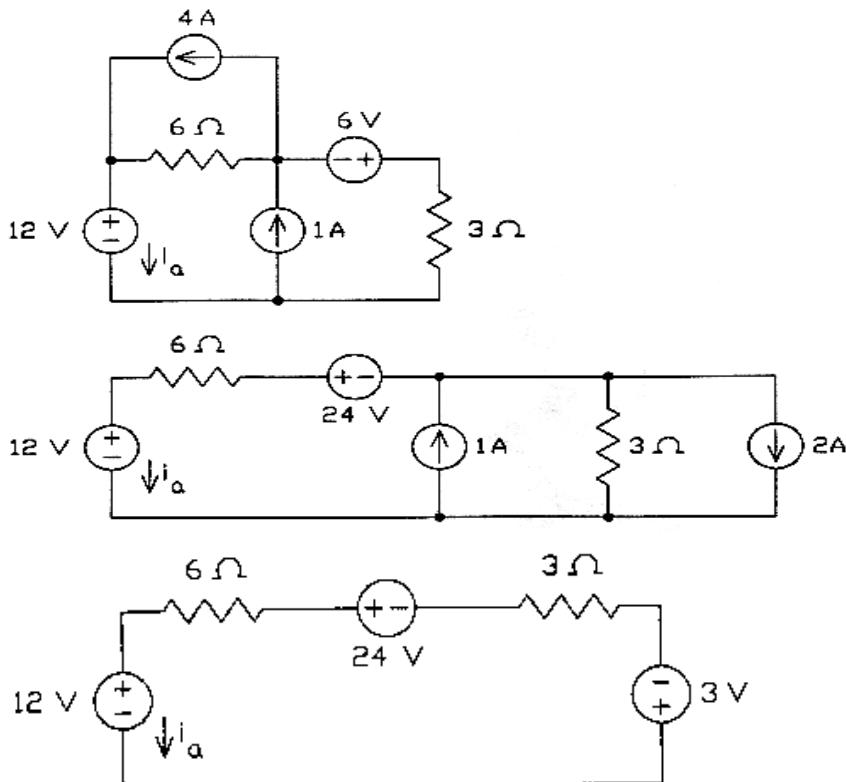
$$\therefore i_a = 375 \mu\text{A}$$

P5.3-6



Section 5-4 Superposition

P5.4-1



$$-12 - 6i_a + 24 - 3i_a - 3 = 0 \Rightarrow i_a = 1 \text{ A}$$

P5.4-2 Consider 6A source only (open 9A source)



From current divider:

$$v_1 / 20 = 6 \left[\frac{15}{15 + 30} \right] \Rightarrow v_1 = 40 \text{ V}$$

Consider 9A source only (open 6A source)

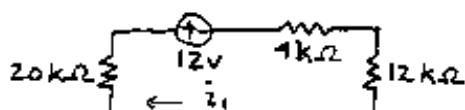


Current divider

$$v_2 / 20 = 9 \left[\frac{10}{10 + 35} \right] \Rightarrow v_2 = 40 \text{ V}$$

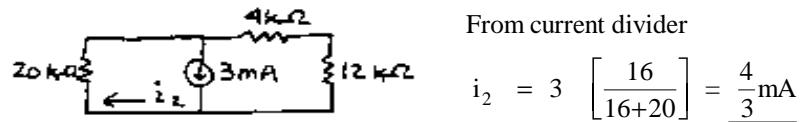
$$\therefore v = v_1 + v_2 = 40 + 40 = 80 \text{ V}$$

P5.4-3 Consider 12V source only (open both current sources)

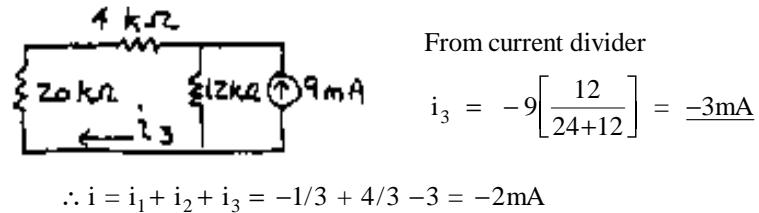


$$\begin{aligned} \text{KVL a: } & 20i_1 + 12 + 4i_1 + 12i_1 = 0 \\ \Rightarrow & i_1 = -1/3 \text{ mA} \end{aligned}$$

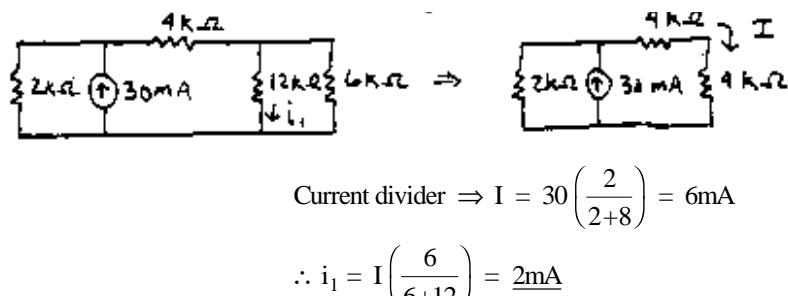
Consider 12mA source only (short 12V and open 6mA sources)



Consider 9mA source only (short 12V and open 12mA sources)

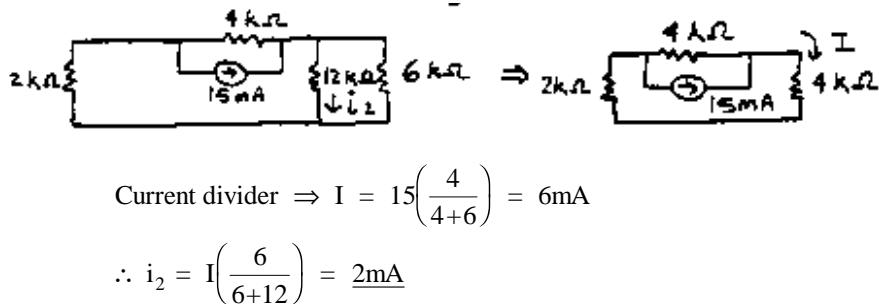


P5.4-4 Consider 30mA source only (open 15mA and short 60V sources)

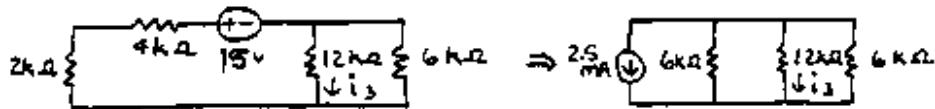


Consider 15mA source only (open 30mA source and short 60V source)

Continued



Consider 15V source only (open both current sources)

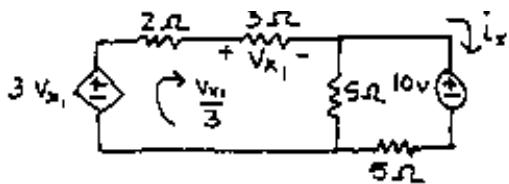


From current divider

$$i_3 = -2.5 \left(\frac{6/6}{6/6+12} \right) = -10 \left(\frac{3}{3+12} \right) = \underline{\underline{-0.5 \text{ mA}}}$$

$$\therefore i = i_1 + i_2 + i_3 = 2 + 2 - 0.5 = 3.5 \text{ mA}$$

P5.4-5 Consider 10V source only (open 4A source)



KVL 1st mesh a:

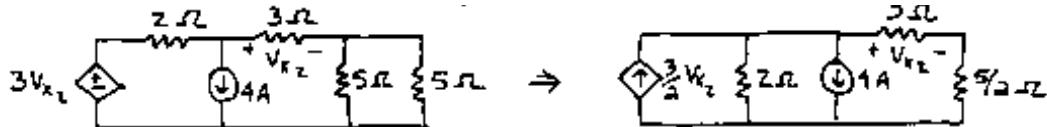
$$-3v_{x_1} + 5\left(\frac{v_{x_1}}{3}\right) + 5\left(\frac{v_{x_1}}{3} - i_x\right) = 0$$

$$\Rightarrow v_{x_1} = 15i_x \quad (1)$$

KVL 2nd mesh a: $5(i_x - v_{x_1}/3) + 10 + 5i_x = 0 \quad (2)$

Solving (1) and (2) simultaneously $\Rightarrow v_{x_1} = 10V$

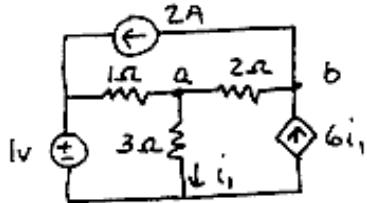
Consider 4A source only (short 10V source)



Using current divider: $\frac{v_{x_2}}{3} = \left(3/2v_{x_2} - 4\right) \left(\frac{2}{2+3+5/2} \right) \Rightarrow v_{x_2} = 16V$

$\therefore v_x = v_{x_1} + v_{x_2} = 10 + 16 = 26V$

P5.4-6



KCL at b: $i + 6i_1 - 2 = 0$

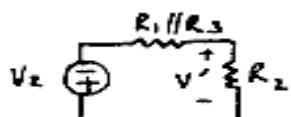
$$\Rightarrow i_1 = 1/3 - 1/6 i \quad (1)$$

KVL around left lower mesh:

$$1(i_1 + i) + 3i_1 - 1 = 0 \quad (2)$$

Plugging (1) into (2) $\Rightarrow i = -1A$

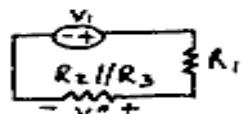
P5.4-7 Consider v_2 source only



Voltage divider: $v' = -v_2 \left[\frac{R_2}{R_2 + R_1 \parallel R_3} \right]$

$$v' = -v_2 \left[\frac{R_2(R_1 + R_3)}{R_1 R_2 + R_1 R_3 + R_2 R_3} \right]$$

Consider v_1 source only



Voltage divider $v'' = v_1 \left[\frac{R_2 \parallel R_3}{R_2 \parallel R_3 + R_1} \right]$

$$v'' = v_1 \left[\frac{R_2 R_3}{R_1 R_2 + R_1 R_3 + R_2 R_3} \right]$$

Consider i_1 source only

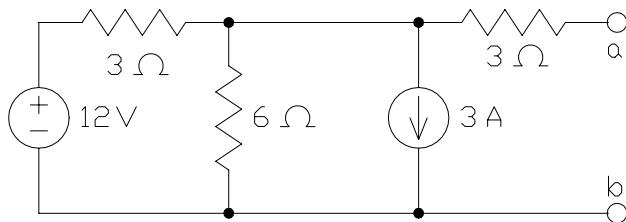


$v''' = 0$ since no current flows through R_2, R_3 and R_1

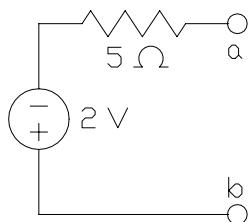
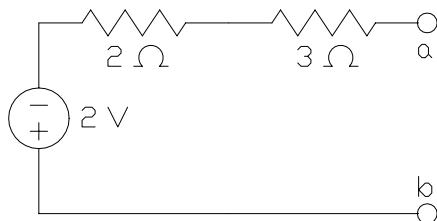
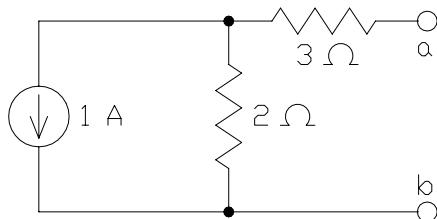
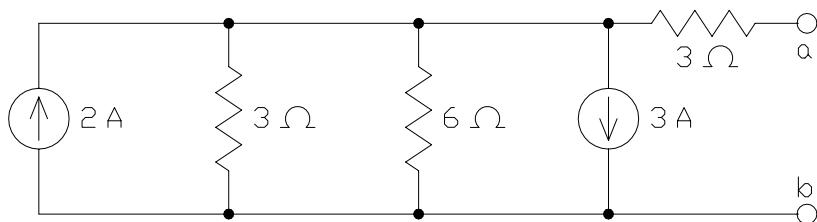
$$\therefore v = v' + v'' + v''' = \frac{v_1 R_2 R_3 - v_2 (R_2 (R_1 + R_3))}{R_1 R_2 + R_1 R_3 + R_2 R_3}$$

Section 5-5: Thèvenin's Theorem

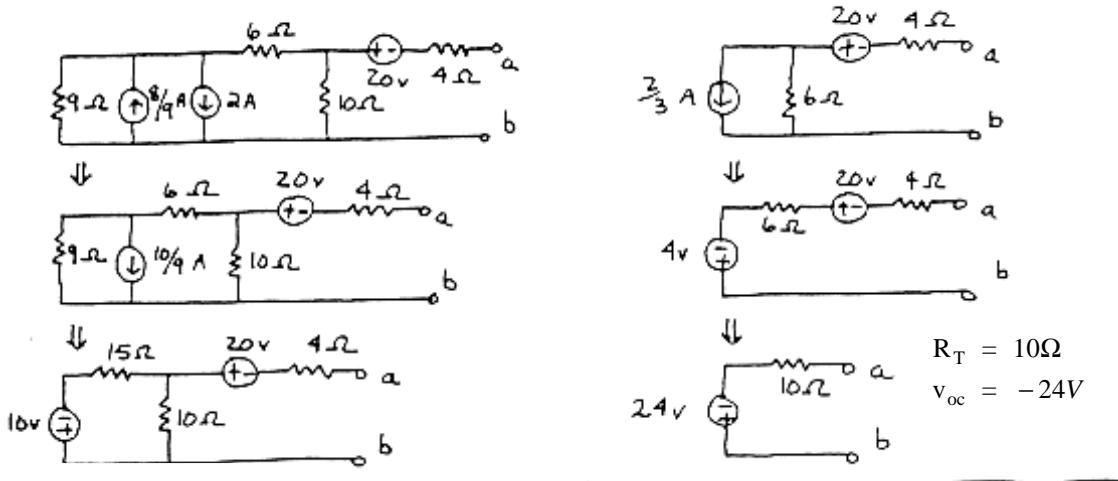
Ex 5.5-1



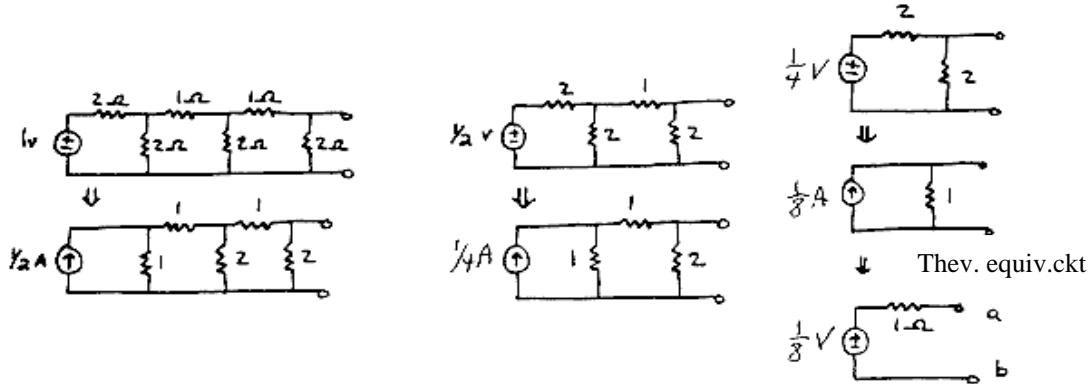
(a)



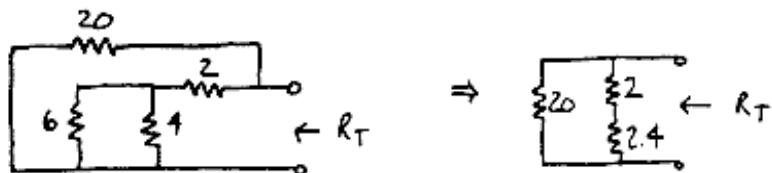
P5.5-2 Use source transformations



P5.5-3 Use source transformations



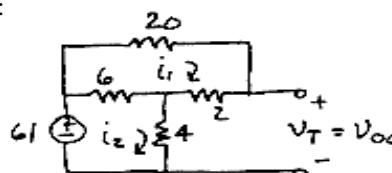
P5.5-4 Find R_T :



$$R_T = \frac{20(2+2.4)}{20+2+2.4} = 3.61\Omega$$

Continued

Find v_T :



$$v_T = 2i_1 + 4i_2$$

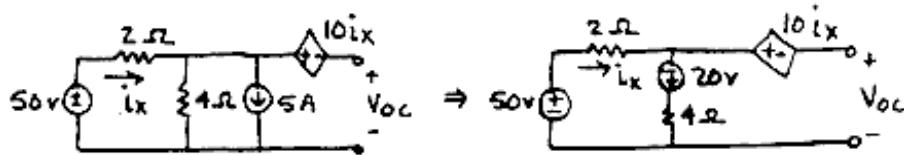
$$\text{mesh } i_1 : 28i_1 - 6i_2 = 0 \quad (1)$$

$$\text{mesh } i_2 : -6i_1 + 10i_2 - 61 = 0 \quad (2)$$

Solving (1) & (2) yields: $i_1 = 1.5A$, $i_2 = 7A$

$$\therefore v_T = 3 + 28 = 31V$$

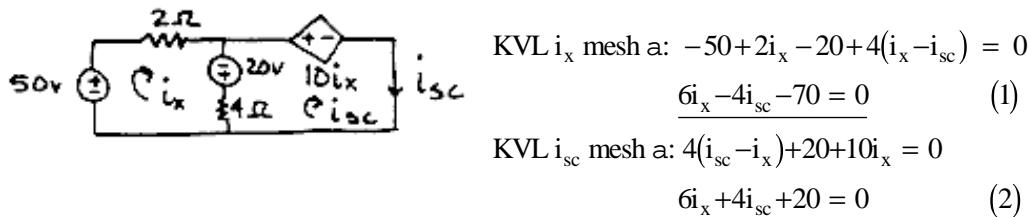
P5.5-5 Find v_{oc}



$$\text{KVL around 1st mesh } a: -50 + 2i_x - 20 + 4i_x = 0 \Rightarrow i_x = \frac{70}{6} \text{ A}$$

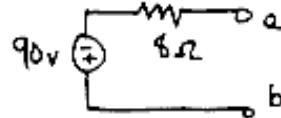
$$\begin{aligned} \text{KVL around 2nd mesh } a: -4i_x + 20 + 10i_x + v_{oc} &= 0 \\ \Rightarrow v_{oc} &= -90 \text{ V} \end{aligned}$$

Find i_{sc}



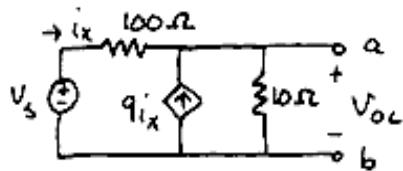
$$\text{Solving (1) and (2) simultaneously } \Rightarrow i_{sc} = -\frac{45}{4} \text{ A}$$

$$\therefore R_T = \frac{v_{oc}}{i_{sc}} = 8 \Omega \quad \text{Thev. equiv. ckt:}$$



P5.5-6

For v_{oc} :



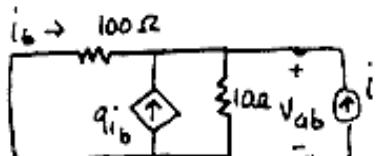
$$i_x = \frac{v_s - v_{oc}}{100}$$

KCL at terminal a:

$$\frac{1}{100}(v_{oc} - v_s) - 9 \left[\frac{1}{100}(v_s - v_{oc}) \right] + \frac{1}{10}v_{oc} = 0$$

$$\Rightarrow v_{oc} = \frac{1}{2}v_s$$

Use current source at a-b to find R_T :

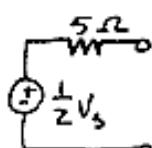


$$i_b = -\frac{v_{ab}}{100}$$

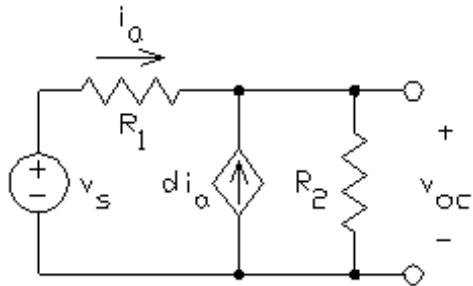
$$\text{KCL: } \frac{1}{100}v_{ab} - 9 \left[\frac{1}{100}(-v_{ab}) \right] + \frac{1}{10}v_{ab} - i = 0$$

$$\Rightarrow i = \frac{1}{5}v_{ab} \quad \therefore R_T = \frac{v_{ab}}{i} = 5 \Omega$$

So Thev. equiv.



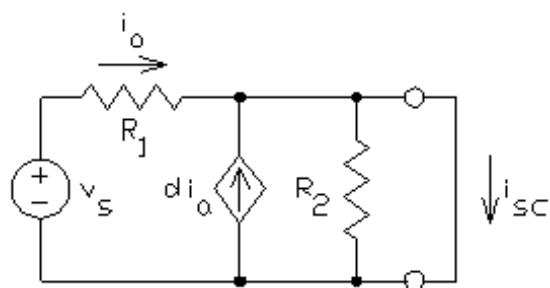
P5.5-7



$$v_s + R_1 i_a + (d+1)R_2 i_a = 0$$

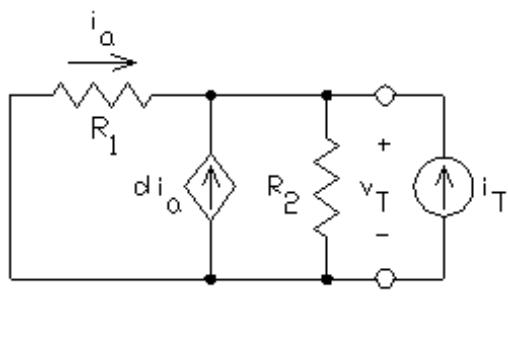
$$i_a = \frac{v_s}{R_1 + (d+1)R_2}$$

$$v_{oc} = \frac{(d+1)R_2 v_s}{R_1 + (d+1)R_2}$$



$$i_a = \frac{v_s}{R_1}$$

$$i_{sc} = (d+1)i_a = \frac{(d+1)v_s}{R_1}$$



$$-i_a - d i_a + \frac{v_T}{R_2} - i_T = 0$$

$$R_1 i_a = -v_T$$

$$i_T = (d+1) \frac{v_T}{R_1} + \frac{v_T}{R_2} = \frac{R_2(d+1) + R_1}{R_1 R_2}$$

$$R_t = \frac{v_T}{i_T} = \frac{R_1 R_2}{R_1 + (d+1)R_2}$$

(b) Let $R_1 = R_2 = 1 \text{ k}\Omega$. Then

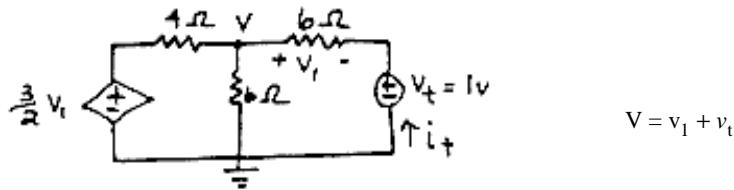
$$625 \Omega = R_t = \frac{1000}{d+2} \Rightarrow d = \frac{1000}{625} - 2 = -0.4 \text{ A/A}$$

and

$$5 = \frac{(d+1)v_s}{d+2} \Rightarrow v_s = \frac{-0.4 + 2}{-0.4 + 1} 5 = 13.33 \text{ V}$$

P5.5-8

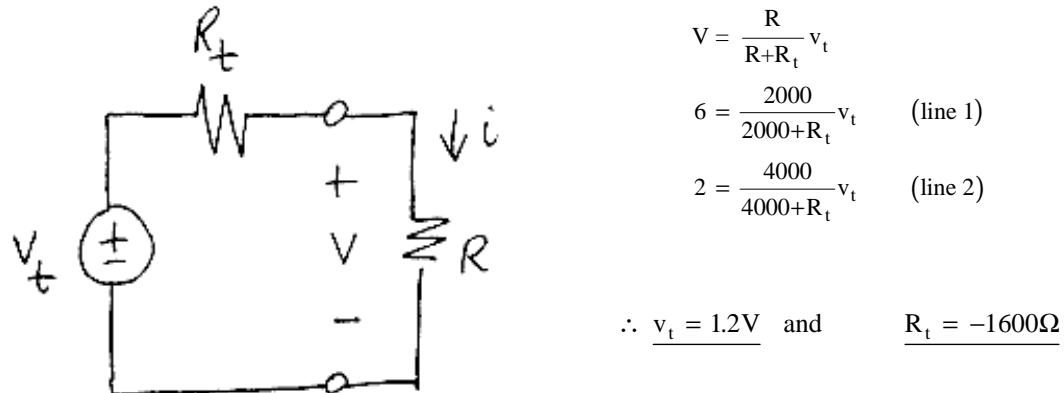
Since no independent sources $v_{oc} = i_{sc} = 0 \therefore \text{apply test source}$



$$\text{KCL at } V: \left(\frac{V - 3v_1}{2}\right) \Big/ 4 + \left(\frac{V}{6} + \frac{v_1}{6}\right) = 0 \quad \& \text{ with } V = v_1 + 1 \text{ so } v_1 = -2A$$

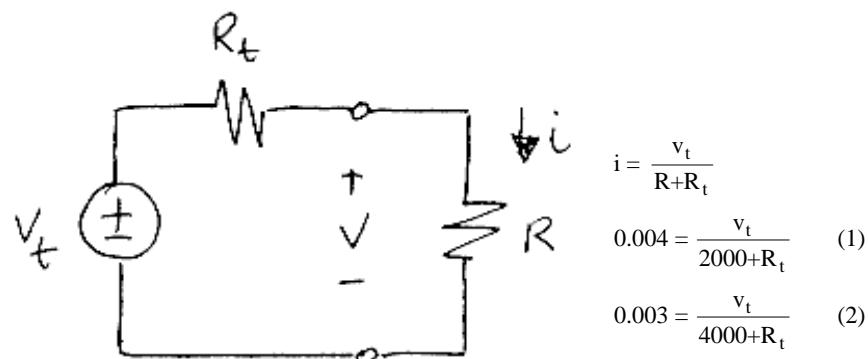
$$\text{now } i_t = -\frac{v_1}{6} = \frac{1}{3}A \quad \therefore R_T = \frac{v_t}{i_t} = \frac{1}{\frac{1}{3}} = 3\Omega$$

P5.5-9



$$\text{When } R = 8000, V = \frac{8000}{8000 - 1600} 1.2 = 1.5V$$

P5.5-10



$$\text{so } v_t = 24V \text{ and } R_t = 4000\Omega$$

$$(a) \quad 0.002 = \frac{24}{R+4000} \Rightarrow R = 8000\Omega$$

$$(b) \quad \text{when } R = 0 \text{ then } i = \frac{24}{4000} = 6 \text{ mA}$$

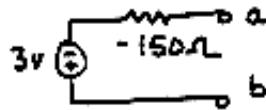
P5.5-11

From the graph, when $v_{ab} = v = 0 \Rightarrow i = i_{sc} = 20 \text{ mA}$

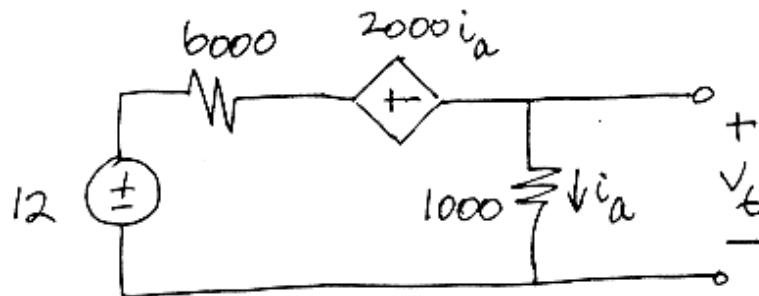
when $i = 0 \Rightarrow v = v_{oc} = -3 \text{ V}$

$$\therefore R_T = \frac{v_{oc}}{i_{sc}} = \frac{-3\text{V}}{20 \text{ mA}} = -15 \text{ k}\Omega = -150 \text{ }\Omega$$

Thev. equiv. ckt \Rightarrow



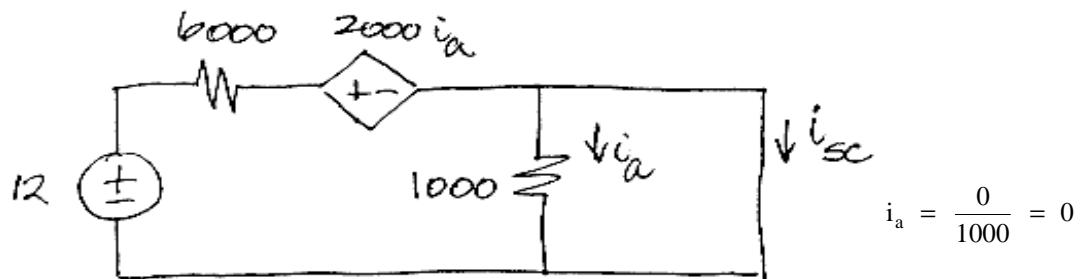
P5.5-12



$$-12 + 6000i_a + 2000i_a + 1000i_a = 0$$

$$i_a = 4/3000 \text{ A}$$

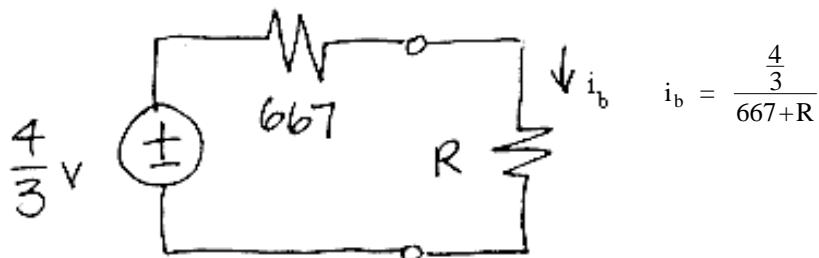
$$v_t = 1000i_a = \frac{4}{3} \text{ V}$$



$$i_a = \frac{0}{1000} = 0$$

$$-12 + 6000i_{sc} = 0 \Rightarrow i_{sc} = 2 \text{ mA}$$

$$R_t = \frac{v_t}{i_{sc}} = \frac{\frac{4}{3}}{.002} = 667 \text{ }\Omega$$

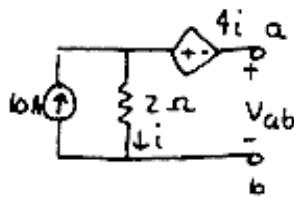


$$\therefore i_b = 0.002 \text{ requires } R = \frac{\frac{4}{3}}{0.002} - 667 = 0$$

P5.5-13

1) disconnect R_L

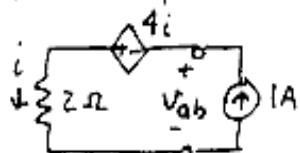
open circuit a - b



$$\text{KVL: } -v_{ab} - 4i + 2i = 0, i = 10\text{A}$$

$$\Rightarrow v_T = v_{ab} = -2i = \underline{-20\text{V}}$$

2) set independent source = 0 and place 1A source at a - b

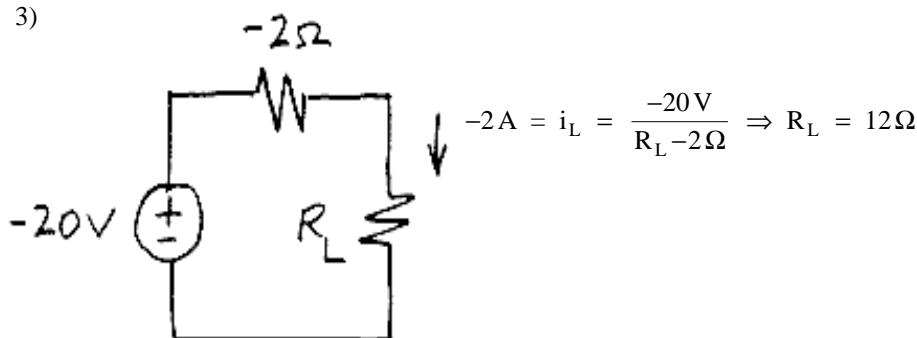


$$\text{KVL: } -v_{ab} - 4i + 2i = 0, i = 1\text{A}$$

$$\Rightarrow v_{ab} = -2\text{A}$$

$$\therefore R_T = v_{ab}/1\text{A} = \underline{-2\Omega}$$

3)



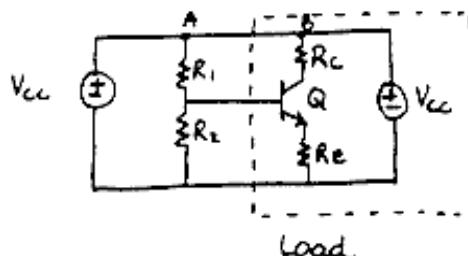
$$-2\text{A} = i_L = \frac{-20\text{V}}{R_L - 2\Omega} \Rightarrow R_L = 12\Omega$$

P5.5-14

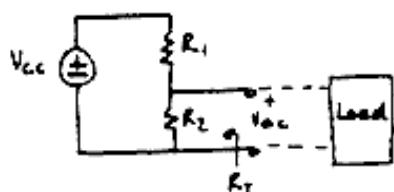
When the terminals of the boxes are open-circuited, no current flows in Box A, but the resistor in Box B dissipates 1 watt. Box B is therefore warmer than Box A. If you short the terminals of each box, the resistor in Box A will draw 1 amp and dissipate 1 watt. The resistor in Box B will be shorted, draw no current, and dissipate no power. Then Box A will warm up and Box B will cool off.

P5.5-15

Redraw ckt as:



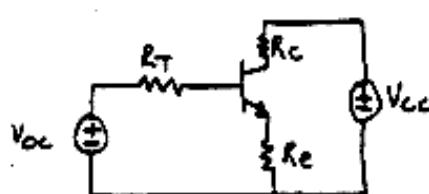
Since points A & B are at same potential, virtually no current exists between A-B ∴ open ckt.



Find R_T : kill v_{cc} source $\Rightarrow R_T = R_1 \parallel R_2 = R_1 R_2 / R_1 + R_2$

Find v_{oc} : voltage divider $v_{oc} = v_{cc} \left(\frac{R_2}{R_1 + R_2} \right)$

can replace above ckt as :

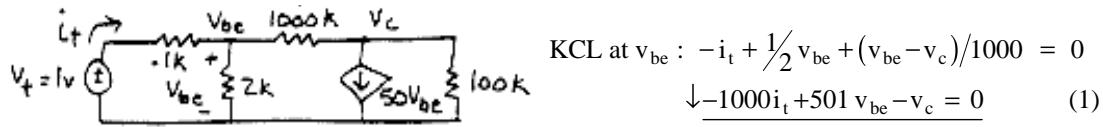


where $v_{oc} = v_{cc} \left(\frac{R_2}{R_1 + R_2} \right)$

$$R_T = \frac{R_2 R_1}{R_2 + R_1}$$

P5.5-16

(a) Since there are no independent sources, apply test source



$$\text{KCL at } v_c : (v_c - v_{be})/1000 + 50v_{be} + v_c/100 = 0$$

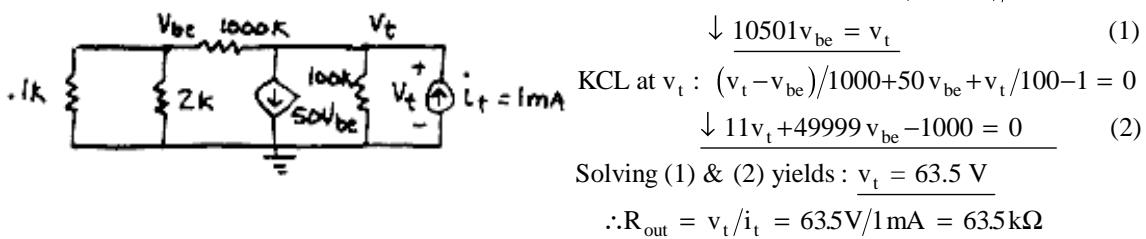
$$\downarrow 11v_c + 50000v_{be} = 0 \quad (2)$$

$$\text{also } (1 - v_{be})/1 = i_t \quad (3)$$

Solving (1), (2), & (3) simultaneously yields $i_t = 3.35 \text{ mA}$

$$\therefore R_{IN} = \frac{v_t}{i_t} = \frac{1 \text{ V}}{3.35 \text{ mA}} = .299 \text{ k}\Omega = 299 \text{ }\Omega$$

(b) Apply test source



P5.5-17

When $0 < V < V_p$, it works as a pure resistor

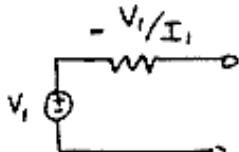
$$\text{so } R = V_p/I_p \quad V_{oc} = 0$$

When $V_p < V < V_m$, it is linear but shows negative resistance characteristic

$$\Rightarrow V_{oc} = V_{oc}|_{I=0} = V_1$$

$$R = \frac{V_{oc}}{I_{sc}} = -\frac{V_1}{I_1}$$

$$-\frac{V_p}{I_p}$$

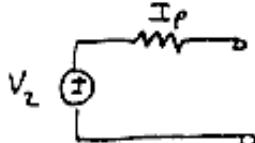


When $V_m < V < V_f$, it is linear

$$\text{so } V_{oc} = V|_{I=0} = V_2$$

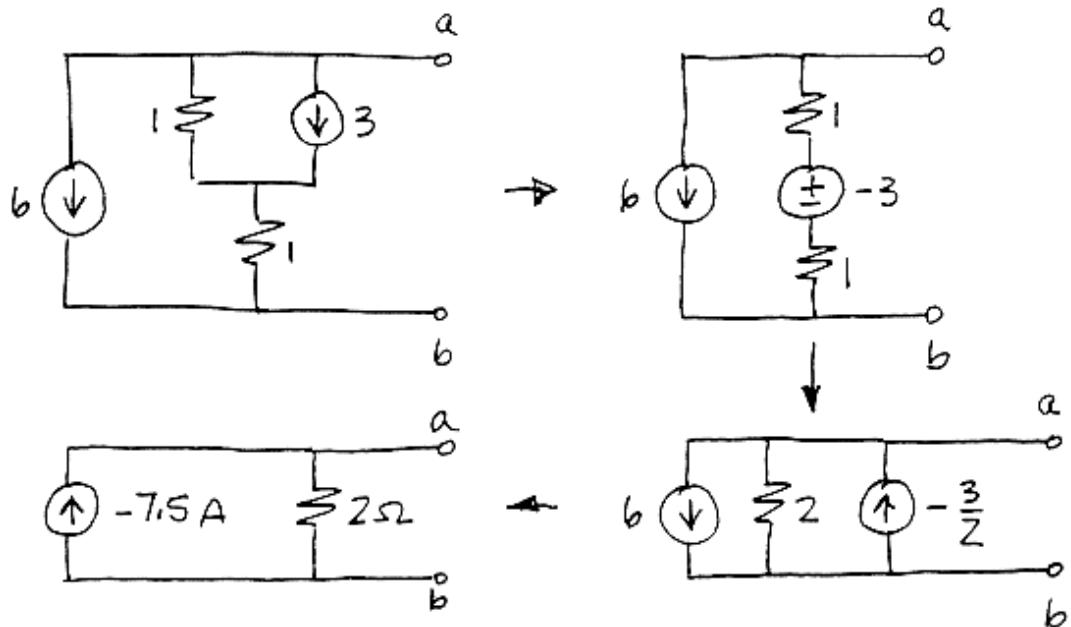
$$R = \frac{V_{oc}}{I_{sc}} = \frac{V_f - V_2}{I_p}$$

$$-\frac{V_f - V_2}{I_p}$$

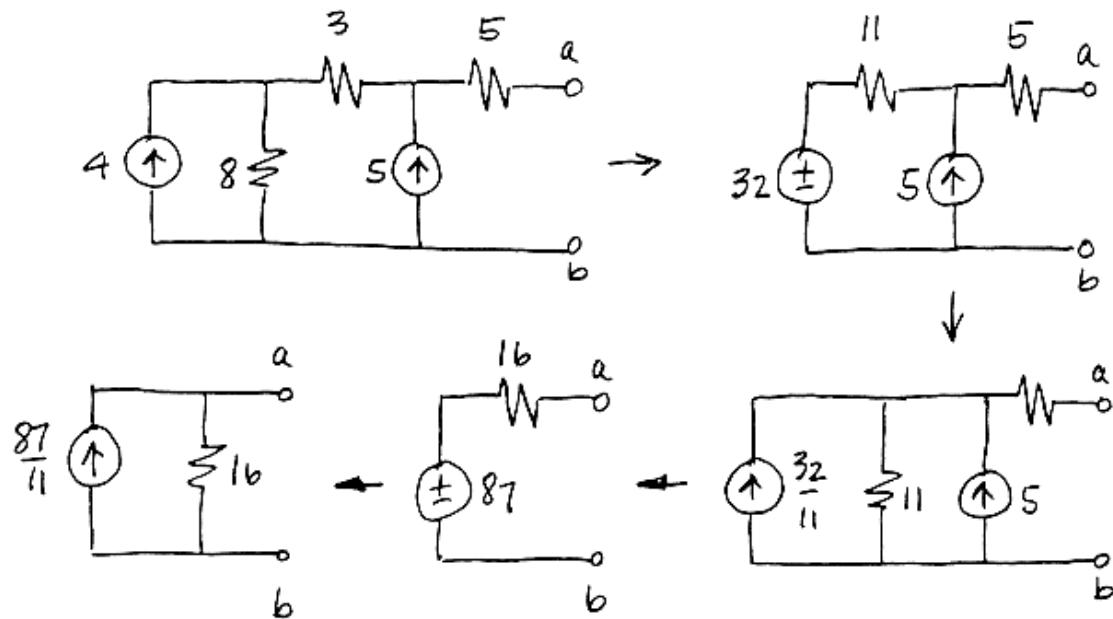


Section 5-6: Norton's Theorem

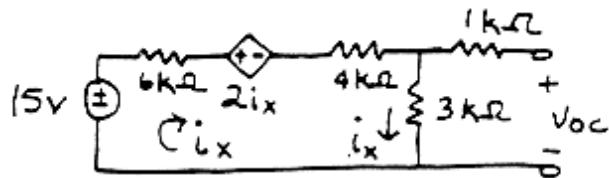
P5.6-1



P5.6-2



P5.6-3 Find v_{oc}

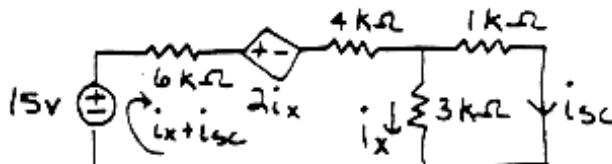


$$\text{KVL } \alpha i_x: -15 + i_x(6+4) + 2i_x + 3i_x = 0$$

$$\Rightarrow i_x = 1\text{mA}$$

$$\therefore v_{oc} = 3i_x = 3\text{V}$$

Find i_{sc}



$$\text{KVL } \alpha i_x + i_{sc}: -15 + 2i_x + 10(i_x + i_{sc}) + 3i_x = 0$$

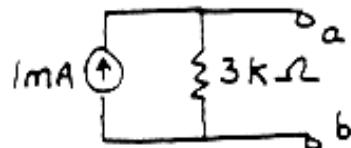
$$\Rightarrow -15 + 15i_x + 10i_{sc} = 0 \quad (1)$$

$$\text{KVL } \alpha i_{sc}: -3i_x + i_{sc} = 0 \quad (2)$$

Solving (1) & (2) simultaneously yields: $i_{sc} = 1\text{mA}$

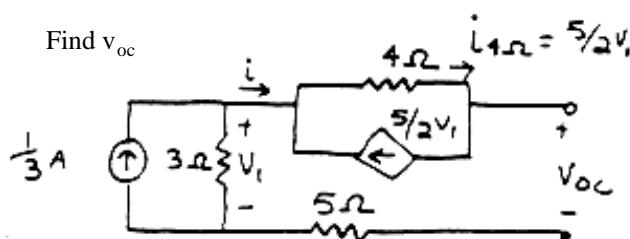
$$\therefore R_T = \frac{v_{oc}}{i_{sc}} = \frac{3}{1} = 3\text{k}\Omega$$

Norton equiv. ckt.



P5.6-4

Find v_{oc}



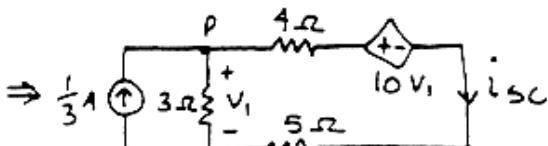
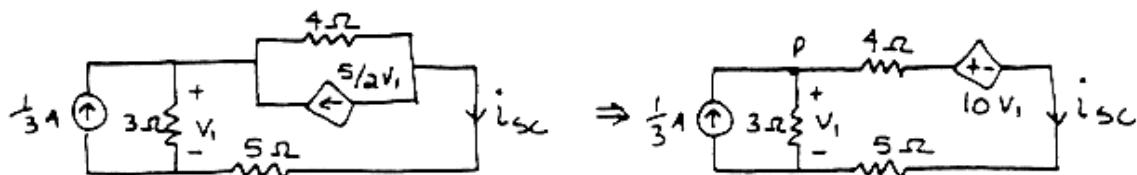
by inspection $i = 0$

from left mesh: $v_1 = 3(1/3) = 1\text{V}$

from KVL a: $-v_1 + 4i_{4\Omega} + v_{oc} = 0$

$$\Rightarrow v_{oc} = v_1 - 4(5/2 v_1) = -9\text{V}$$

Find i_{sc}



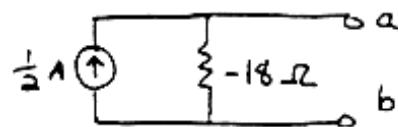
from KVL a: $-v_1 + 4i_{sc} + 10v_1 + 5i_{sc} = 0$

$$\Rightarrow 9v_1 + 9i_{sc} = 0 \quad (1)$$

from KCL at P: $-\frac{1}{3} + \frac{v_1}{3} + i_{sc} = 0 \quad (2)$ (1) & (2) yields

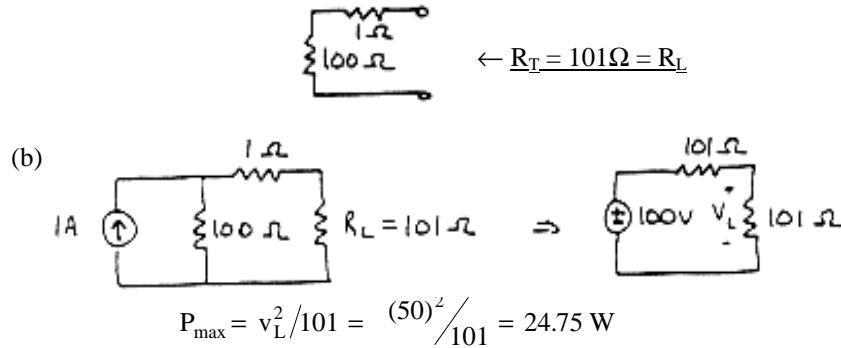
$$\therefore R_T = \frac{v_{oc}}{i_{sc}} = \frac{-9}{1/2} = -18\Omega$$

Norton equiv. ckt: $i_{sc} = \frac{1}{2}\text{A}$



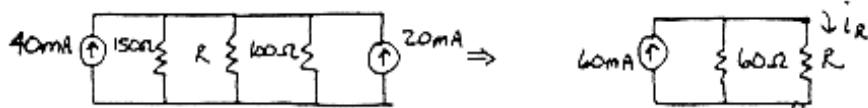
Section 5-7: Maximum Power Transfer

P5.7-1



P5.7-2

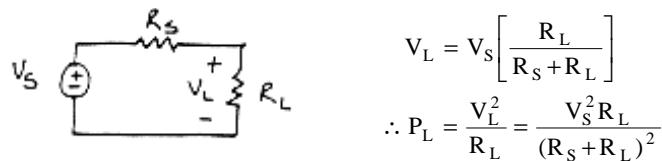
(a) Use source transformations to reduce ckt.



Norton equiv. where $R_T = 60\Omega$ \therefore want $R_L = 60\Omega$

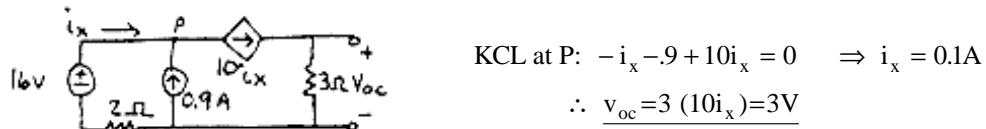
$$(b) P_{\max} = i_R^2(R) = (30)^2(60) = 54,000\mu \text{ W} = 54 \text{ mW}$$

P5.7-3

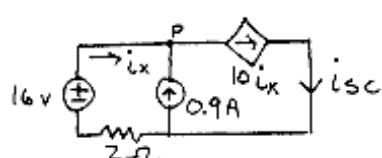


By inspection, P_L is max when you vary R_S to get the smallest denominator. \therefore set $R_S = 0$

P5.7-4 Find R_T using $R_T = v_{oc}/i_{sc}$. First find v_{oc} :



Find i_{sc}

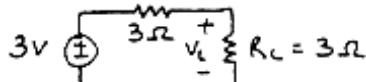


KCL at P: $-i_x - 0.9 + 10i_x = 0 \Rightarrow i_x = 0.1 \text{ A}$

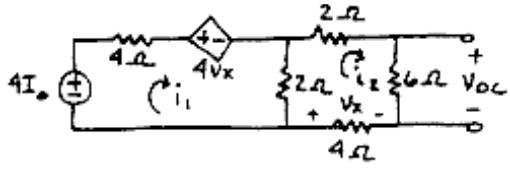
$$\therefore i_{sc} = 10i_x = 1 \text{ A}$$

$$P_{L_{\max}} = \frac{V_L^2}{R_L} = \frac{(1.5)^2}{3} = 0.75 \text{ W}$$

$$\therefore R_T = v_{oc}/i_{sc} = 3\Omega = R_L \quad \text{for max power}$$



P5.7-5 (a) For max power $R_L = R_T$. First find v_{oc} :



$$\text{KVL at } i_1: -4I_0 + 4i_1 - 4v_x + 2(i_1 - i_2) = 0 \\ \Rightarrow 6i_1 - 2i_2 + 4v_x - 4I_0 = 0 \quad (1)$$

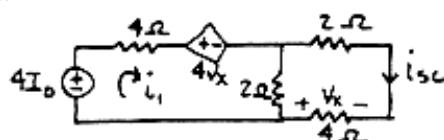
$$\text{KVL at } i_2: 2i_2 + 6i_2 - v_x + 2(i_2 - i_1) = 0 \\ \Rightarrow -2i_1 + 10i_2 - v_x = 0 \quad (2)$$

$$\text{also } v_x = 4i_2 \quad (3)$$

$$\text{and } v_{oc} = 6i_2 \quad (4)$$

Solving (1), (2), (3), & (4) yields $v_{oc} = I_0$

Find i_{sc}



$$\text{KVL at } i_1: -4I_0 + 4i_1 + 4v_x + 2(i_1 - i_{sc}) = 0 \\ \Rightarrow 6i_1 - 2i_{sc} + 4v_x - 4I_0 = 0 \quad (1)$$

$$\text{KVL at } i_x: 2i_{sc} + 4i_{sc} + 2(i_{sc} - i_1) = 0 \\ \Rightarrow -2i_1 + 8i_{sc} = 0 \quad (2)$$

$$\text{also } v_x = -4i_{sc} \quad (3)$$

$$\text{Solving (1), (2), & (3) yields } i_{sc} = \frac{2}{3} I_0 \quad \therefore R_T = \frac{v_{oc}}{i_{sc}} = \frac{3}{2} \Omega = R_L$$

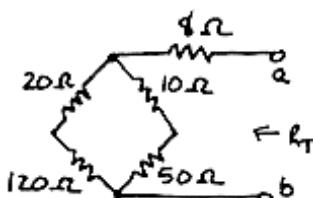
$$(b) P_{L_{\max}} = 54 \text{ W} = \frac{(v_{oc}/2)^2}{R_L} = I_0^2 / 6$$

$$\Rightarrow I_0 = 18 \text{ A}$$

P5.7-6

$$P_{\max} = v_T^2 / 4R_T$$

Find $R_T \Rightarrow$ kill i source

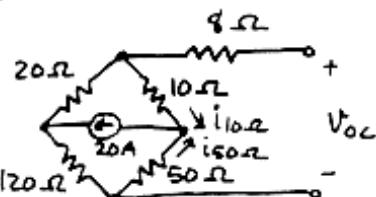


$$R_T = 8 + (20+120)\parallel(10+50) \\ = 50 \Omega$$

find v_{oc} :

$$i_{10\Omega} = \frac{120 + 50}{120+50+20+10} 20 \text{ A} \\ = 17 \text{ A}$$

$$\therefore v_{10\Omega} = 10(17) = 170 \text{ V}$$

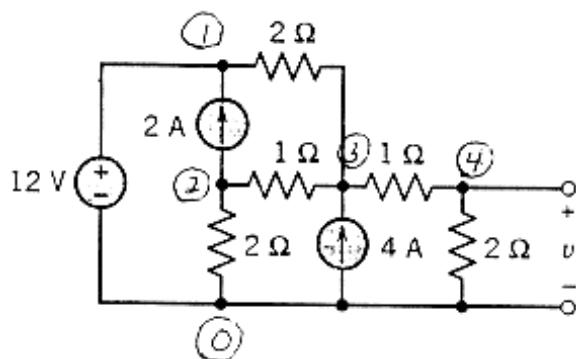


$$v_{50\Omega} = 50(17 - 20) = -150 \text{ V} \Rightarrow v_{oc} = v_{10\Omega} + v_{50\Omega} = 170 - 150 = 20 \text{ V}$$

$$\therefore P_{\max} = \underline{20^2} = 2 \text{ W}$$

PSpice Problems

SP 5-1



Input file

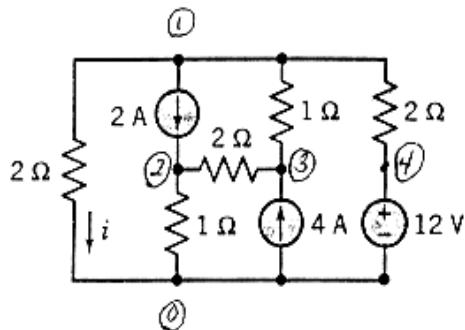
```

V1    1      0      dc      12
I2    2      1      dc      2
R3    2      0      2
R4    2      3      1
R5    1      3      2
R6    3      4      1
I7    0      3      dc      4
R8    4      0      2
.dc   V1 12 12 1
.print dc V (4)
.END

```

result $V(4) = v = 4.952E+00V$

SP 5-2



Input file

```

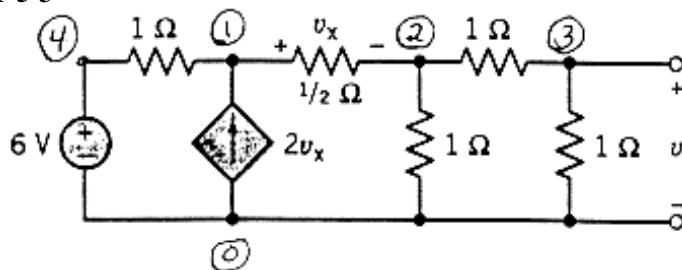
R1    1      0      2
I2    1      2      dc      2
R3    2      0      1
R4    2      3      2
R5    1      3      1
I6    0      3      dc      4
R7    1      4      2
V8    4      0      dc      12
.dc   V8 12 12 1
.print dc I(R1)
.END

```

result

$i = I(R1) = 3.000E+00A$

SP 5-3



result

$v = v(3) = 1.714E+00V$

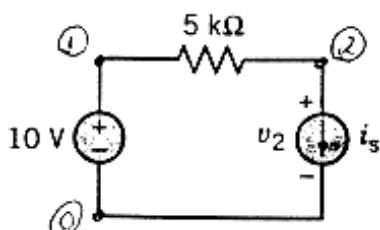
Input file

```

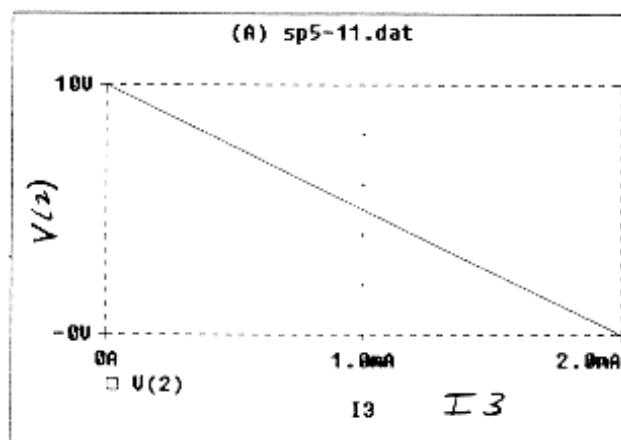
V1    4      0      dc      6
G2    0      1      1      2
R3    1      2      500m
R4    1      4      1
R5    2      0      1
R6    2      3      1
R7    3      0      1
.dc   V1 6 6 1
.print dc V(3)
.END

```

SP 5.4



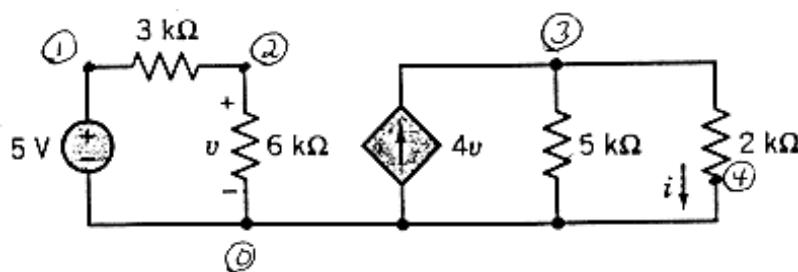
Probe result



Input file

```
V1    1      0      dc      10
R2    1      2      5k
I3    2      0      dc      1m
.dc   I3 0  2e-3   0.2e-3
.probe V(2)
.end
```

SP 5.5

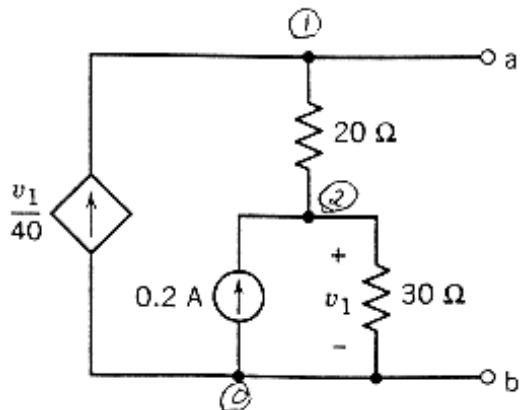


Input file

```
V1    1      0      dc      5
R1    1      2      3k
R2    2      0      6k
G4    0      3      2      0      4
R3    3      0      5k
R4    3      0      2k
.dc   V1  5  5  1
.print  dc  I(R4)
.end
```

result $i = I(R4) = 9.524E+00A$

SP 5-6



Input file

```

G1    0      1      2      0      25m
R2    1      2      20
I3    0      2      dc     0.2
R4    2      0      30
.tf   V(1)   I3
.end

```

result

answer: $V_1 = V_{oc} = V_T = 36V$

NODE	VOLTAGE	NODE	VOLTAGE
(1)	<u>36.0000</u>	(2)	24.0000

$R_{TH} = \text{OUTPUT RESISTANCE AT } V(1)=2.000E+02\Omega$

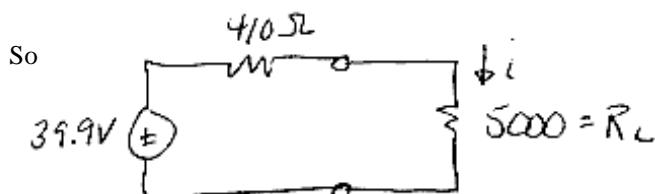
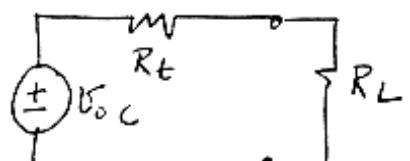
Verification Problems

VP 5-1 Evaluating data

$$\text{Case 1: } R_L = 0\Omega ; i = I_{sc} = 97.2\text{mA} = \frac{V_{oc}}{R_t} \quad (1)$$

$$\text{Case 2: } R_L = 500\Omega ; i = 43.8\text{mA} = \frac{V_{oc}}{R_t + 500} \quad (2)$$

Solving 1+2 yields $R_t = 410\Omega , V_{oc} = 39.9\text{V}$



When $R_L = 5000\Omega$

$$i = \frac{V_{oc}}{R_t + R_L} = 7.37\text{mA}$$

not 16.5mA as recorded \therefore the data is inconsistent.

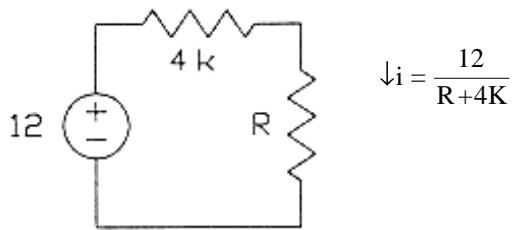
VP 5-2

$$V_{oc} = 12 \text{ V} \quad (\text{line 1 of the table})$$

$$I_{sc} = 3 \text{ mA} \quad (\text{line 3 of the table})$$

so

$$R_{TH} = V_{oc} / I_{sc} = 4 \text{ k}\Omega$$



$$\downarrow i = \frac{12}{R+4\text{k}}$$

Hence the circuit can be simplified as shown above right. (Check:

$$\frac{12}{10\text{k}+4\text{k}} = 0.857 \text{ mA}$$

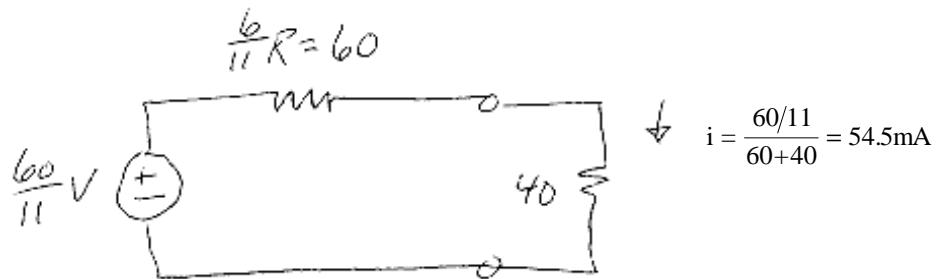
as shown in line 2 of the table.)

When $i = 1 \text{ mA}$ is required

$$1 \text{ mA} = \frac{12}{R+4\text{k}} \Rightarrow R = \frac{12}{1 \text{ mA}} - 4\text{k} = 8\text{k}\Omega$$

I agree with my lab partner's claim that $R = 8000$ causes $i = 1 \text{ mA}$.

VP 5-3



$$\downarrow i = \frac{60/11}{60+40} = 54.5 \text{ mA}$$

The measurement is consistent with the prelab calculations.

Design Problems

DP 5-1 The equation of representing the straight line in Figure DP 5-1b is $v = -R_t i + v_{oc}$. That is, the slope of the line is equal to -1 times the Thevenin resistance and the "v - intercept" is equal to the open circuit voltage. Therefore: $R_t = -\frac{0-5}{0.008-0} = 625 \Omega$ and $v_{oc} = 5 \text{ V}$.

Try $R_1 = R_2 = 1 \text{ k}\Omega$. ($R_1 \parallel R_2$ must be smaller than $R_t = 625 \Omega$.) Then

$$5 = \frac{R_2}{R_1 + R_2} v_s = \frac{1}{2} v_s \Rightarrow v_s = 10 \text{ V}$$

and

$$625 = R_3 + \frac{R_1 R_2}{R_1 + R_2} = R_3 + 500 \Rightarrow R_3 = 125 \Omega$$

Now v_s , R_1 , R_2 and R_3 have all been specified so the design is complete.

DP 5-2 The equation of representing the straight line in Figure DP 5-2b is $v = -R_t i + v_{oc}$. That is, the slope of the line is equal to -1 times the Thevenin resistance and the "v - intercept" is equal to the open circuit voltage. Therefore: $R_t = -\frac{0-(-3)}{-0.006-0} = 500 \Omega$ and $v_{oc} = -3 \text{ V}$.

From the circuit we calculate

$$R_t = \frac{R_3(R_1 + R_2)}{R_1 + R_2 + R_3} \text{ and } v_{oc} = -\frac{R_1 R_3}{R_1 + R_2 + R_3} i_s$$

so

$$500 \Omega = \frac{R_3(R_1 + R_2)}{R_1 + R_2 + R_3} \text{ and } 3 \text{ V} = -\frac{R_1 R_3}{R_1 + R_2 + R_3} i_s$$

Try $R_3 = 1 \text{ k}\Omega$ and $R_1 + R_2 = 1 \text{ k}\Omega$. Then $R_t = 500 \Omega$ and

$$-3 = -\frac{1000R_1}{2000} i_s = \frac{R_1}{2} i_s \Rightarrow 6 = R_1 i_s$$

This equation can be satisfied by taking $R_1 = 600 \Omega$ and $i_s = 10 \text{ mA}$. Finally, $R_2 = 1 \text{ k}\Omega - 400 \Omega = 600 \Omega$. Now i_s , R_1 , R_2 and R_3 have all been specified so the design is complete.

DP 5-3 The slope of the graph is positive so the Thevenin resistance is negative. This would require

$$R_3 + \frac{R_1 R_2}{R_1 + R_2} < 0, \text{ which is not possible since } R_1, R_2 \text{ and } R_3 \text{ will all be non-negative.}$$

Is it not possible to specify values of v_s , R_1 , R_2 and R_3 that cause the current i and the voltage v in Figure DP 5-3a to satisfy the relationship described by the graph in Figure DP 5-3b.

DP 5-4 The equation of representing the straight line in Figure DP 5-4b is $v = -R_t i + v_{oc}$. That is, the slope of the line is equal to the Thevenin impedance and the "v - intercept" is equal to the open circuit voltage. Therefore: $R_t = -\frac{-5-0}{0-0.008} = -625 \Omega$ and $v_{oc} = -5 \text{ V}$.

The open circuit voltage, v_{oc} , the short circuit current, i_{sc} , and the Thevenin resistance, R_t , of this circuit are given by

$$v_{oc} = \frac{R_2(d+1)}{R_1 + (d+1)R_2} v_s,$$

$$i_{sc} = \frac{(d+1)}{R_1} v_s$$

and

$$R_t = \frac{R_1 R_2}{R_1 + (d+1)R_2}$$

Let $R_1 = R_2 = 1 \text{ k}\Omega$. Then

$$-625 \Omega = R_t = \frac{1000}{d+2} \Rightarrow d = \frac{1000}{-625} - 2 = -3.6 \text{ A/A}$$

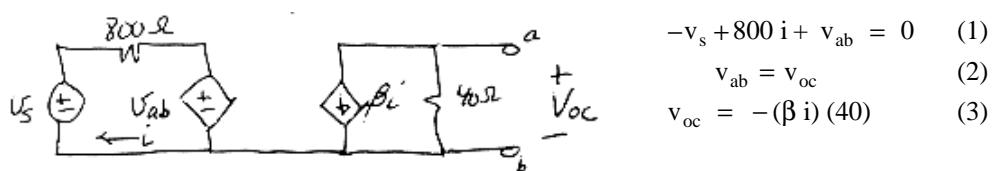
and

$$-5 = \frac{(d+1)v_s}{d+2} \Rightarrow v_s = \frac{-3.6+2}{-3.6+1}(-5) = -3.077 \text{ V}$$

Now v_s , R_1 , R_2 and d have all been specified so the design is complete.

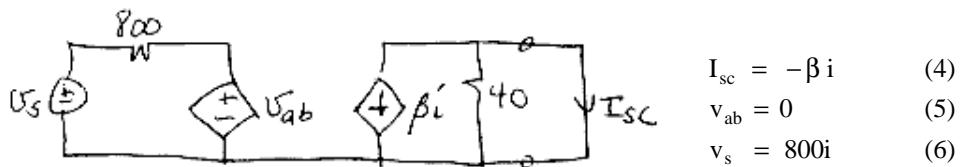
DP 5-5 a) Find Thev. equiv.

v_{oc} :



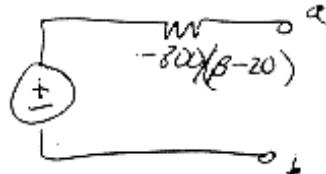
Solving eqs. (1) - (3) yields $v_{oc} = v_s / \left(1 - \frac{20}{\beta}\right)$

I_{sc} :



Solving eqs. (4) - (6) yields $I_{sc} = -\beta v_s / 800$

so $R_t = \frac{v_{oc}}{I_{sc}} = \frac{-800}{\beta - 20}$ and Thev. equiv. $\frac{v_s}{\left(1 - \frac{20}{\beta}\right)}$



b) $R_t = R_L = 400 = \frac{-800}{\beta-20} \Rightarrow \underline{\beta = 18}$

c) Max power to R_L , largest v_{oc} , largest v_s , smallest R_t

$$P_L = \frac{(V_L)^2}{R_L} \quad (7)$$

$$\text{and } V_L = \frac{400}{R_{\text{total}}} v_{oc} \quad (8)$$

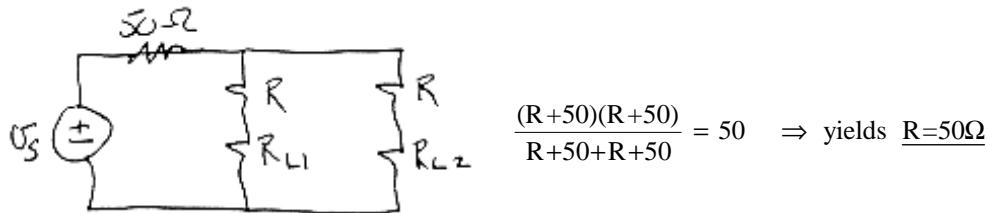
with

$$R_{\text{total}} = \frac{-800}{\beta-20} + 400 \text{ (a) yields } \underline{\beta = \pm 18}$$

d) delivering large amounts of power could melt antenna.

DP 5-6 Max power to load : $R_L = R_t = 50\Omega$

But split power equally ($R_{L1} = R_{L2} = 50\Omega$)



$$\frac{(R+50)(R+50)}{R+50+R+50} = 50 \Rightarrow \text{yields } \underline{R=50\Omega}$$