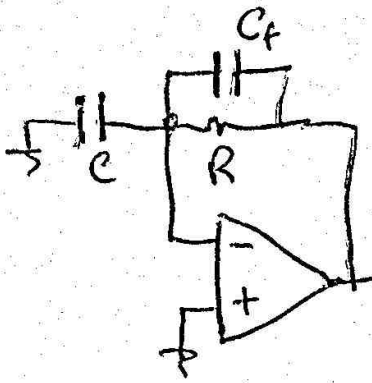


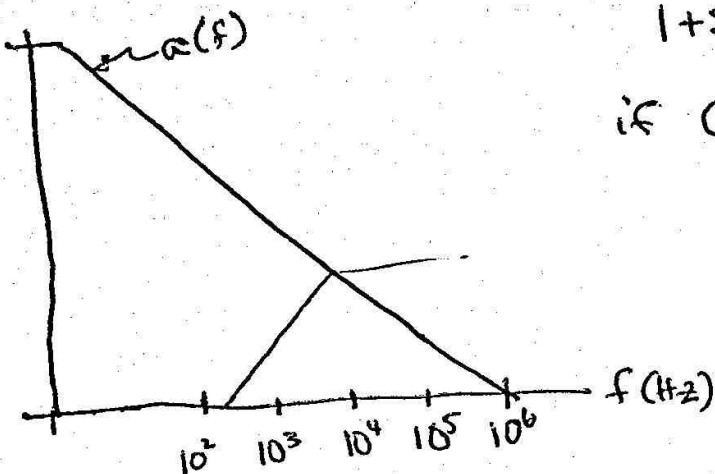
8-14



$$Z_2 = \frac{1}{sC_F} \parallel R = \frac{R/sC_F}{R + 1/sC_F} = \frac{R}{1 + sC_F R}$$

$$\beta = \frac{1/sC}{Z_2 + 1/sC} = \frac{1}{sC Z_2 + 1}$$

$$= \frac{1}{sCR + 1} = \frac{1 + sC_F R}{1 + sR(C + C_F)}$$



if $C \gg C_F$

$$f_z \approx \frac{1}{2\pi RC} = 202 \text{ Hz}$$

↳ zero of $1/\beta$

For $\phi_m = 45^\circ$ select

f_p to lie at the geometrical average of f_z and f_z

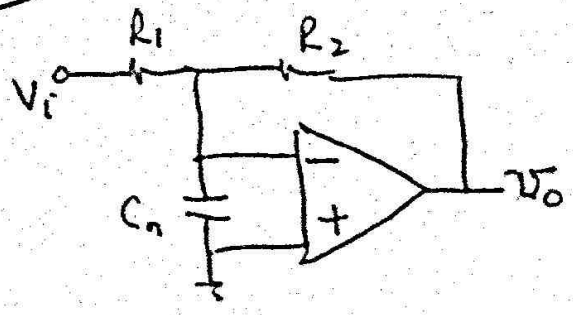
$$f_p = \sqrt{(202.23)(10^6)} = 14.22 \text{ kHz}$$

$$C_F = \frac{1}{2\pi R f_p} = \frac{1}{2\pi(78.7 \text{ k}\Omega)(14.22 \text{ kHz})}$$

$$C_F = 0.142 \text{ nF} \leftarrow \ll 10 \text{ nF as expected.}$$

$$\underline{\underline{142 \text{ pF}}}$$

8-18



(a)

$$C_n = C_d + \frac{1}{2} C_c + C_{ext} = 16 \text{ pF}$$

$$GBP = 20 \text{ MHz}$$

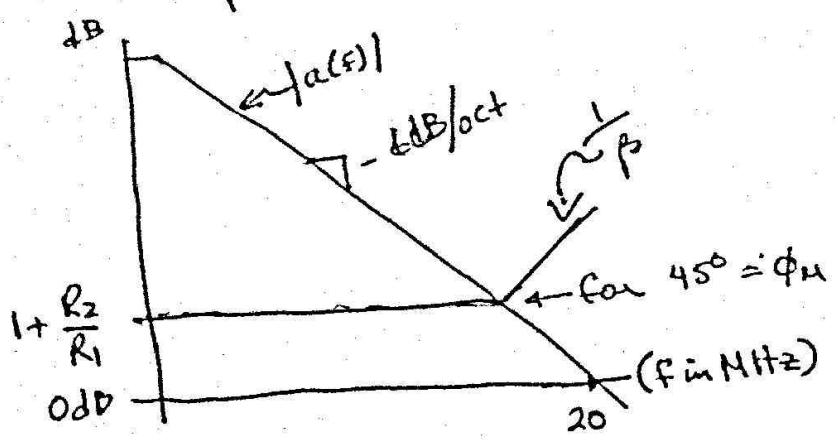
$$\beta = \frac{Z_1}{Z_1 + R_2} ; Z_1 = \frac{R_1 / sC_n}{R_1 + 1/sC_n}$$

$$Z_1 = \frac{R_1}{1 + sC_n R_1} \Rightarrow \beta = \frac{R_1 / (1 + sC_n R_1)}{R_2 + R_1 / (1 + sC_n R_1)}$$

$$\beta = \frac{R_1}{R_2 + sC_n R_1 R_2 + R_1} = \frac{R_1}{R_1 + R_2} \frac{1}{1 + sC_n (R_1 || R_2)}$$

$$= \frac{1}{1 + R_2/R_1} \frac{1}{1 + sC_n (R_1 || R_2)}$$

$$\frac{1}{\beta} = \left(1 + \frac{R_2}{R_1}\right) \left[1 + sC_n (R_1 || R_2)\right]$$



$$1 + \frac{R_2}{R_1} = 2 = 6 \text{ dB}$$

∴ place fz at 10 MHz

$$\frac{1}{2\pi C_n (R_1 || R_2)} = 10 \text{ MHz} \Rightarrow R_1 || R_2 = \frac{1}{2} R_1 = \frac{1}{2} R_2$$

$$\frac{1}{2} R_1 = \frac{1}{2\pi (16 \text{ pF}) (10^7 \text{ Hz})}$$

$$R_1 = R_2 = 2 (995 \Omega) = \boxed{1990 \Omega}$$

(b) for $\phi_M = 60^\circ$

$$\tan\left(\frac{f_z}{f_c}\right) = 30^\circ \rightarrow f_z = \frac{f_c}{\tan 30^\circ} = 17.3 \text{ MHz}$$

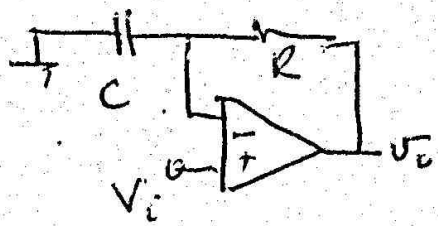
$$\boxed{R_1 = R_2 = 3445.9 \Omega}$$

advantages: requires no additional parts, less noise

advantages : requires no additional parts , less noise

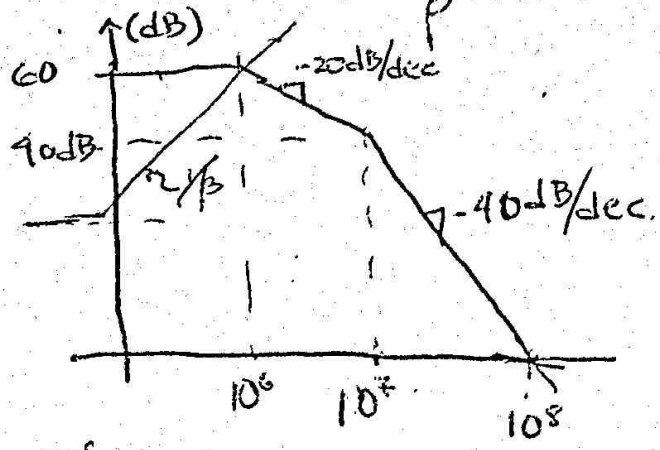
disadvantages : more output current flows through feedback network. Large effect of Thevenin resistance.

8.27 (a) $a_0 = 10^3 \text{ V/V}$, $f_1 = 10^6 \text{ Hz}$, $f_2 = 10^7 \text{ Hz}$



$$\beta = \frac{1/sC}{R + 1/sC} = \frac{1}{sCR + 1}$$

$$\frac{1}{\beta} = sCR + 1$$



If the $1/\beta$ curve is made to cross the a curve at the first pole then the system is stable with $\phi_m = 45^\circ$

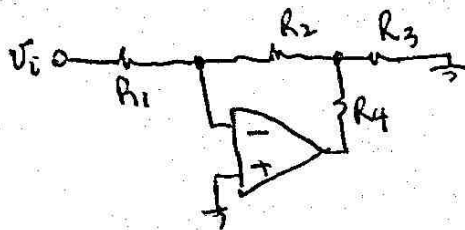
(b) Select

$$\frac{1}{2\pi RC} = 10^3 \text{ Hz} \quad (\text{3 decades below } f_1)$$

Let $C = 1 \mu\text{F}$, $R = \frac{1}{2\pi(10^3)(10^{-6})} = \frac{1 \text{ M}\Omega}{2\pi}$

(c) SS BW is the freq. at which $1/\beta$ crosses the "a" curve. So $\text{BW} = f_1 = 1 \text{ MHz}$

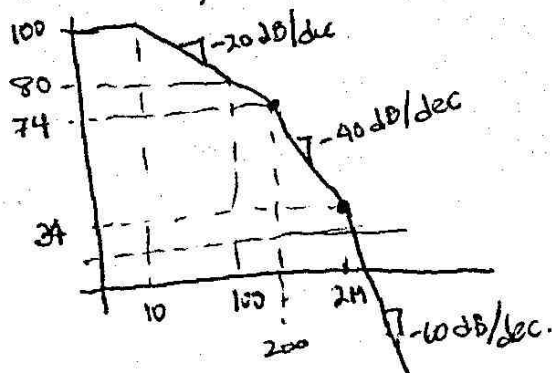
8-34



$$(a) \quad \beta = \frac{R_1}{R_1 + R_2} \frac{(R_1 + R_2) \parallel R_3}{R_4 + (R_1 + R_2) \parallel R_3}$$

$$= \frac{1}{2} \frac{200 \parallel 10}{200 \parallel 10 + 100} \approx \frac{1}{23}$$

$a_0 = 10^5 \text{ V/V}$, $f_1 = 10 \text{ kHz}$, $f_2 = 200 \text{ kHz}$, $f_3 = 2 \text{ MHz}$

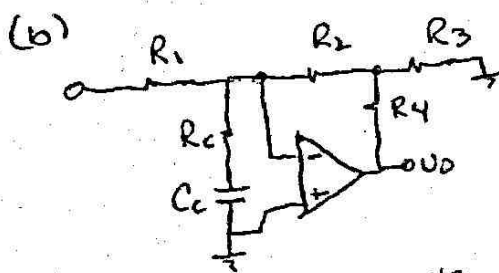


$$\frac{1}{\beta} = 27.2 \text{ dB}$$

(crosses |a| a bit above 2 MHz)

$$\therefore \phi < -\tan^{-1}\left(\frac{2 \text{ M}}{10 \text{ K}}\right) - \tan^{-1}\left(\frac{2 \text{ M}}{200 \text{ K}}\right) - \tan^{-1}\left(\frac{2}{2}\right) = -219^\circ$$

and the amp. is unstable.

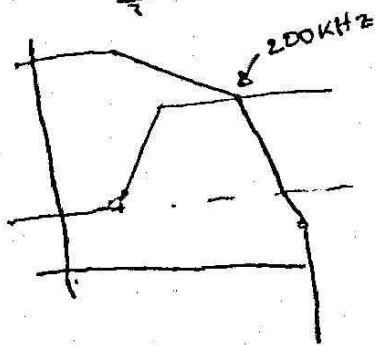


let $R_a = R_1 \parallel R_c$

$$\beta = \frac{R_a}{R_a + R_2} \frac{(R_a + R_2) \parallel R_3}{R_4 + (R_a + R_2) \parallel R_3}$$

Gain at 2nd pole is $\sim 100 \text{ dB}$
 $- 29 \text{ dB}$
 $\frac{71 \text{ dB}}$

$$\approx 3550 = \frac{1}{\beta}$$



$$\frac{1}{\beta} = \left(1 + \frac{R_2}{R_a}\right) \left[\frac{10 \parallel (100 + R_a)}{100 + 10 \parallel (100 + R_a)} \right]^{-1} \approx 3550$$

Let $10 \parallel (100 + R_a) \approx 10$

$$\left(1 + \frac{100}{R_a}\right) \frac{10}{10} \approx 3550$$

$$R_a = \left(\frac{10(3550) - 1}{100} \right)^{-1} = 311.2$$

$$311.2 = 100 \text{ K} \parallel R_c =$$

$$R_c = \frac{1}{\frac{1}{311} - \frac{1}{100}} = \boxed{312.5 \Omega}$$

$$C_c = \frac{5}{\pi (312.2)(200 \text{ kHz})} = \boxed{25.5 \text{ nF}}$$

$$8.38 \quad (a) \quad Z_1 = \frac{1}{sC_c} \parallel R = \frac{R/sC_c}{R + 1/sC_c} = \frac{R}{sCR + 1}$$

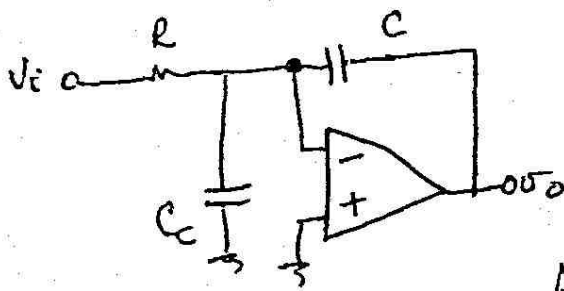
$$\beta = \frac{Z_1}{Z_1 + 1/sC} = \frac{R/(1+sCR)}{\frac{R}{1+sCR} + 1/sC} = \frac{sCR}{sCR + 1 + sC_c R}$$

$$\beta = \frac{sCR}{1 + s(C_c + C)R}$$

$$\frac{1}{\beta} (f = \omega) = 1 + \frac{C_c}{C} = 5 \rightarrow \boxed{C_c = 4C = 160 \text{ pF}}$$

$$(b) \quad f_x = \frac{GBP}{5} = 16 \text{ MHz}$$

Circuit behaves as a single-pole opamp. up to 16 MHz



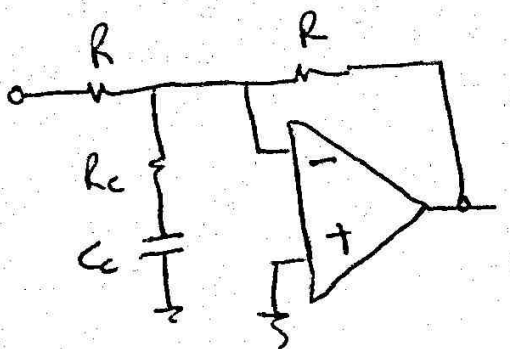
$$A_{ideal} = \frac{-1}{sCR}$$

Above $\frac{1}{\beta}$ is for non-inv. version of the circuit

$$A(s) = \frac{-1}{sCR} \frac{1}{1 + j(f/f_x)}$$

8.40

$GBP = 80\text{MHz}$, $\beta_{max} = 0.2\text{V/V}$, $\frac{1}{\beta_{max}} = 5\text{V/V}$



$f_x = \frac{80\text{MHz}}{5} = 16\text{MHz}$

$R_c = \frac{R}{5 - (1+1)}$ (eq. 8.30)

$R_c = \frac{R}{3} \rightarrow \text{let } \begin{cases} R_c = 10\text{K} \\ R = 30\text{K} \end{cases}$

$C_c = \frac{5}{\pi R_c f_x}$ (eq. 8.31)
 $= \frac{5}{\pi (10\text{K}\Omega)(16\text{MHz})}$

$C_c = 10\text{pF}$

$A(s) = (-1\text{V/V}) \frac{1}{1 + j \frac{f}{16\text{MHz}}}$
 \uparrow
 f_x