

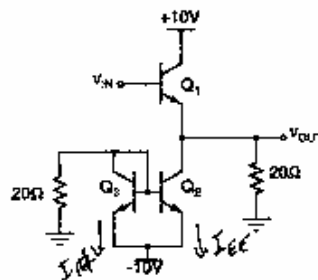
NAME:

KEY

STUDENT NO.:

University of Puerto Rico
 Electrical and Computer Engineering Department
 INEL 4202 - Electronics II - Fall 2002 - Exam 1a - Prof. M. Toledo
 THERE ARE TWO PROBLEMS - BE CLEAR OR LOOSE POINTS

1. The transistors in the circuit shown below are matched. Transistor Q_1 is dissipating an average power of 1W. Find the power being delivered to the load. Assume sinusoidal output. (30 points)



$$I_{EE} = \frac{10V - 0.7V}{20\Omega} = 465mA = I_{EE}$$

dissipated in Q_1, Q_2

Since $P_{DC} = 2V_{CC} I_{EE} = 2(10V)(.465A) = 9.3W$
 and half of that is dissipated in Q_2 ;

$$P_{Q2} = 4.65W$$

$$P_{Load} = P_{DC} - P_{Q2} - P_{Q1} = 9.3W - 4.65W - 1W$$

$$P_{Load} = 3.65W$$

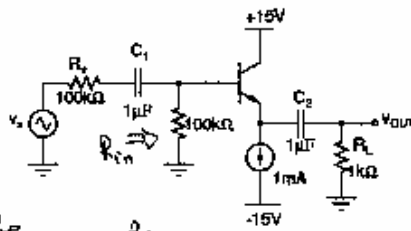
Notice, however, that this implies an efficiency of $\frac{3.65W}{9.3W} \times 100\% \approx 40\%$, which is larger than the max. for a class A amplifier (25%). What happens is that there is clipping at the output ($V_{peak} = \sqrt{2(3.65W)(20\Omega)} \approx 12V$ for an output power of 3.65W, and this is greater than V_{CC}). Still, the power dissipated in the load should be about 3.65W.

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2. For the amplifier shown below, assume $\beta = 100$, $C_{\pi} = 3pF$ and $C_{\mu} = 10pF$. Find

- (a) the midband gain. (10 points)
- (b) the low frequency poles. (20 points)
- (c) the high frequency poles. (40 points)



$$g_m = \frac{1}{25} \frac{I_C}{V_T}$$

$$I_C = 2.5k\Omega$$

(a) $A_{mid} = + \frac{g_m R_c}{1 + g_m R_e} \cdot \frac{R_{in}}{R_s + R_{in}}$

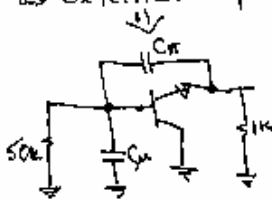
$$= \frac{\frac{1}{25} (1k)}{1 + \frac{1k}{25}} \cdot \frac{100k \parallel (101k + 2.5k)}{100k + 100k \parallel (101k + 2.5k)} = \frac{1}{3} \cdot \frac{40}{41} \approx \boxed{0.325}$$

(b) C_1 sees $R_{eq1} = 100k + 100k \parallel 103.5k \approx 150k \Rightarrow \omega_{L1} = \frac{1}{10^{-6} \times 150k\Omega} \approx \boxed{6.7 \text{ rps}}$

C_2 sees $R_{eq2} = 1k + \frac{r_{\pi} + R_b}{\beta + 1} = 1k + \frac{2.5k + 50k}{101} \approx 1520\Omega$

$\therefore \omega_{L2} = \frac{1}{10^{-6} \times 1520\Omega} = \boxed{658 \text{ rps}}$

(c) Eq. ext. for high-freq. with C_{π} & C_{μ} drawn as external caps.



C_{π} sees $50k \parallel (r_{\pi} + (\beta + 1)R_e) = 50k \parallel 103.5k \approx$

$$\omega_{p1} = \frac{1}{10^{-11} \times 33k\Omega} \approx \boxed{30M \text{ rps}}$$

C_{μ} sees

$V_E = V_B$

$$V_E = 50k i_E + (i_E - \frac{V_E}{25}) 1k$$

$$\frac{V_E}{i_E} = (50k + 1k) / (1 + \frac{1k}{25}) = \frac{51k}{41}$$

$$R_{eq} = r_{\pi} \parallel \frac{V_E}{i_E}$$

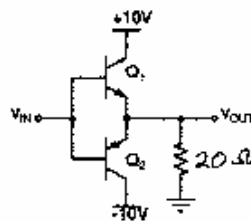
$$= 2.5k \parallel 1244\Omega = 831\Omega \Rightarrow \omega_{pR} = \frac{1}{831\Omega \times 10^{-11} \text{F}} = \boxed{12.0M \text{ rps}}$$

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THERE ARE TWO PROBLEMS - BE CLEAR OR LOOSE POINTS

1. The transistors in the circuit shown below are matched. Transistor Q_1 is dissipating an average power of 0.1w. Find the power being delivered to the load. Assume sinusoidal output. (30 points)



Since

$$P_{Qs} = 2 \times 0.1w = 0.2w = \frac{2}{\pi} \frac{V_{cc}}{R_L} V_{Lpeak} - \frac{V_{Lpeak}^2}{2R_L}$$

$$.2w = \frac{2}{\pi} \frac{10V}{20\Omega} V_{Lpeak} - \frac{1}{40\Omega} V_{Lpeak}^2$$

or

$$V_{Lpeak}^2 - \frac{40}{\pi} V_{Lpeak} + 40 = 0$$

$$V_{Lpeak}^2 - 12.7 V_{Lpeak} + 8 = 0$$

$$V_{Lpeak} = \frac{12.7 \pm \sqrt{12.7^2 - 4(1)(8)}}{2}$$

$$= 6.35 \pm 5.7$$

$$V_{Lpeak} = \cancel{12.05} \text{ or } 12.05$$

\Rightarrow select $0.65V$ because $12.05V$ is larger than $V_{cc} = 10V$.

$$\therefore P_L = \frac{V_{Lpeak}^2}{2R_L} = \frac{(0.65)^2}{2(20)} = \frac{.42}{40} w \approx 10 \text{ mW}$$

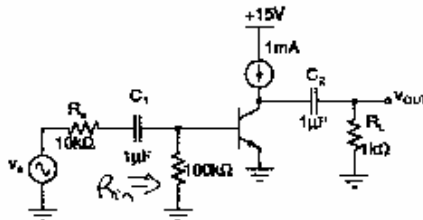
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2. For the amplifier shown below, assume $\beta = 100$, $V_A = 50V$, $C_\mu = 1pF$ and $C_v = 10pF$. Find

- (a) the midband gain. (10 points)
- (b) the low frequency poles. (20 points)
- (c) the high frequency poles. (40 points)

Do not use Miller's theorem.



$$g_m = \frac{I_C}{V_T} = \frac{1mA}{25V}$$

$$r_o = 2.5k\Omega$$

$$(a) A_{mid} = -(g_m R_C) \left(\frac{R_{in}}{R_s + R_{in}} \right) = -\frac{1}{25} (1k) \frac{100k \parallel 2.5k\Omega}{10k\Omega + 100k \parallel 2.5k\Omega}$$

$$A_{mid} = -7.8 \text{ V/V}$$

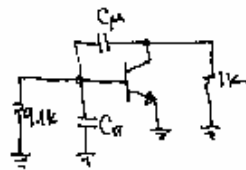
$$(b) R_{eq_{C1}} = 10k + 100k \parallel 2.5k = 12.4k\Omega ; \omega_{p_{C1}} = \frac{1}{10^{-6}F \times 12.4k\Omega} = 80 \text{ rps}$$

$$R_{eq_{C2}} = 1k\Omega + r_o = 1k\Omega + \frac{V_A}{I_C} = 1k\Omega + 50k\Omega = 51k\Omega$$

$$\omega_{p_{C2}} = \frac{1}{10^{-6}F \times 51 \times 10^3} = 19.6 \text{ rps}$$

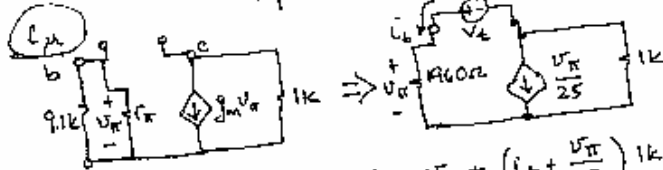
(c)

High freq. CKT
w. C_π & C_μ drawn
as external caps:



$$C_\pi \text{ sees } R_{eq_{B\pi}} = 9.1k \parallel r_\pi = 9.1k \parallel 2.5k = 1960\Omega$$

$$\omega_{p_\pi} = \frac{1}{10^{-6}F \times 1960} = 51 \text{ Mrps}$$



$$v_\pi = 1960 i_b ; v_E = v_\pi + \left(i_b + \frac{v_\pi}{25} \right) 1k$$

$$v_E = 1960 i_b + 1k \left(1 + \frac{1960}{25} \right) i_b$$

$$\frac{v_E}{i_b} = R_{eq_\mu} = 81.4k\Omega \Rightarrow \omega_{p_\mu} = \frac{1}{10^{-12}F \times 81.4k\Omega} = 12.3 \text{ Mrps}$$